

## Apéndice

<b>Hojas de características de componentes</b>		
• Transistor de pequeña señal NPN <b>BC546/548</b>		A.1
• Transistor de pequeña señal NPN <b>2N3903/04</b>		A.6
• Transistor de pequeña señal PNP <b>2N3905/06</b>		A.12
• JFET de canal N <b>2N5457</b>		A.17
• JFET de canal P <b>2N5460/62</b>		A.20
• Amplificador operacional <b>OA741</b>		A.23
• Regulador de tensión ajustable positivo <b>LM117/317</b>		A.29
• Regulador de tensión ajustable negativo <b>LM137/337</b>		A.33
• SIDAC <b>MKP3V129/240</b>		A.36
• SBS <b>MBS4991/2/3</b>		A.38
• SCR <b>2N5060</b>		A.40
• PUT <b>2N6027/28</b>		A.42
• UJT <b>2N2646</b>		A.44
• TRIAC <b>MAC218A</b>		A.45



SILICON PLANAR EPITAXIAL TRANSISTORS

BC546/548

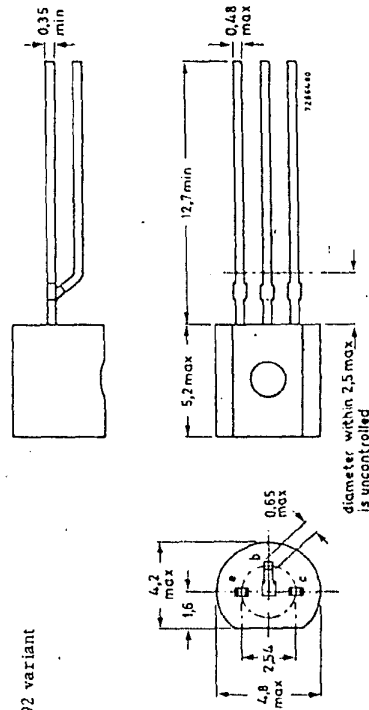
General purpose n-p-n transistors in a plastic TO-92 variant, especially suitable for use in driver stages of audio amplifiers.

QUICK REFERENCE DATA		BC546	BC547	BC548
Collector-emitter voltage ( $V_{BE} = 0$ )	$V_{CES}$ max.	80	50	30
Collector-emitter voltage (open base)	$V_{CEO}$ max.	65	45	30
Collector current (peak value)	$I_{CM}$ max.	200	200	200
Total power dissipation up to $T_{amb} = 25^\circ C$	$P_{tot}$ max.	500	500	500
Junction temperature	$T_j$ max.	150	150	150
Small-signal current gain	$h_{fe}$	$> 125$	$> 125$	$> 125$
		$< 500$	$< 900$	$< 900$
Transition frequency	$f_T$	300	300	300
		typ.	typ.	typ.
Noise figure at $R_S = 2\text{ k}\Omega$	F	2	2	2
		typ.	typ.	typ.

Dimensions in mm

MECHANICAL DATA

TO-92 variant



Accessories: 56356 (cooling clip).

BC546/548

$T_j = 25^\circ\text{C}$  unless otherwise specified

**CHARACTERISTICS**

Collector cut-off current

$I_E = 0; V_{CB} = 30\text{ V}$   
 $I_E = 0; V_{CB} = 30\text{ V}; T_j = 150^\circ\text{C}$

$I_{CBO} < 15\text{ nA}$   
 $I_{CBO} < 5\text{ }\mu\text{A}$

Base-emitter voltage 1)

$I_C = 2\text{ mA}; V_{CE} = 5\text{ V}$

$V_{BE} \text{ typ. } 660\text{ mV}$   
 $580\text{ to }700\text{ mV}$

$I_C = 10\text{ mA}; V_{CE} = 5\text{ V}$

$V_{BE} < 770\text{ mV}$

Saturation voltage 2)

$I_C = 10\text{ mA}; I_B = 0.5\text{ mA}$

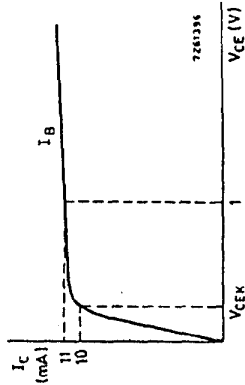
$V_{CEsat} \text{ typ. } 90\text{ mV}$   
 $< 250\text{ mV}$

$I_C = 100\text{ mA}; I_B = 5\text{ mA}$

$V_{BEsat} \text{ typ. } 700\text{ mV}$

Knee voltage

$I_C = 10\text{ mA}; I_B = \text{value for which}$   
 $I_C = 11\text{ mA at } V_{CE} = 1\text{ V}$



Collector capacitance at  $f = 1\text{ MHz}$

$I_E = I_C = 0; V_{CB} = 10\text{ V}$

$C_c \text{ typ. } 2.5\text{ pF}$   
 $< 4.5\text{ pF}$

Emitter capacitance at  $f = 1\text{ MHz}$

$I_C = I_E = 0; V_{EB} = 0.5\text{ V}$

$C_e \text{ typ. } 9\text{ pF}$

Transition frequency at  $f = 35\text{ MHz}$

$I_C = 10\text{ mA}; V_{CE} = 5\text{ V}$

$f_T \text{ typ. } 300\text{ MHz}$

1)  $V_{BE}$  decreases by about  $2\text{ mV}/^\circ\text{C}$  with increasing temperature.

2)  $V_{BEsat}$  decreases by about  $1.7\text{ mV}/^\circ\text{C}$  with increasing temperature.

$T_j = 25^\circ\text{C}$  unless otherwise specified

**CHARACTERISTICS (continued)**

Small signal current gain at  $f = 1\text{ kHz}$

$I_C = 2\text{ mA}; V_{CE} = 5\text{ V}$

$h_{fc} > 125$   
 $< 5000$

Noise figure at  $R_S = 2\text{ k}\Omega$

$I_C = 200\text{ }\mu\text{A}; V_{CE} = 5\text{ V}$

$F \text{ typ. } 2$   
 $< 10$

$f = 1\text{ kHz}; B = 2000\text{ Hz}$

D. C. current gain

$I_C = 10\text{ }\mu\text{A}; V_{CE} = 5\text{ V}$

$h_{FE} \text{ typ. } 90$   
 $> 110$   
 $\text{typ. } 180$   
 $< 220$

$I_C = 2\text{ mA}; V_{CE} = 5\text{ V}$

$h_{FE} > 420$   
 $\text{typ. } 520$   
 $< 800$

h parameters at  $f = 1\text{ kHz}$  (common emitter)

$I_C = 2\text{ mA}; V_{CE} = 5\text{ V}$

$h_{ic} > 1.6\text{ k}\Omega$   
 $\text{typ. } 2.7\text{ k}\Omega$   
 $< 4.5\text{ k}\Omega$

Reverse voltage transfer ratio

$h_{re} \text{ typ. } 1.5$   
 $< 2$

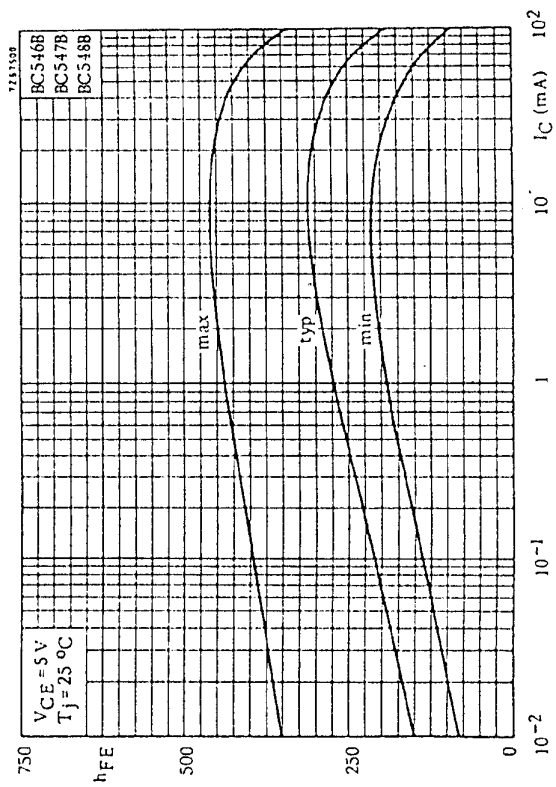
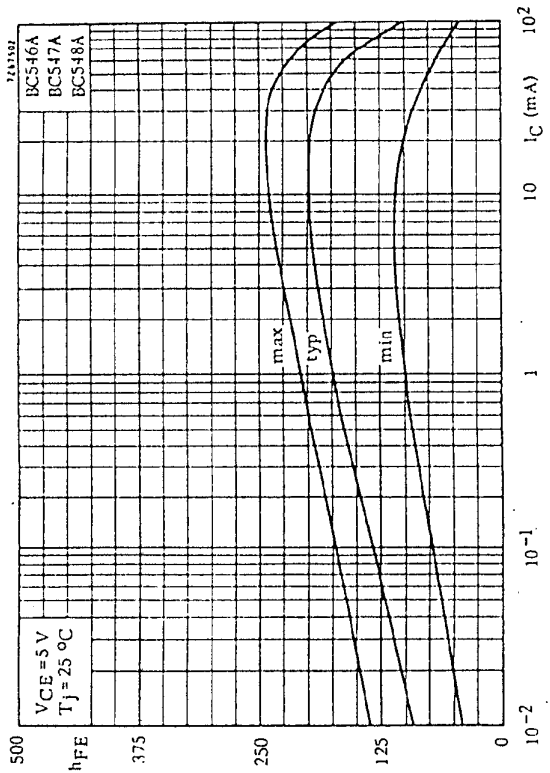
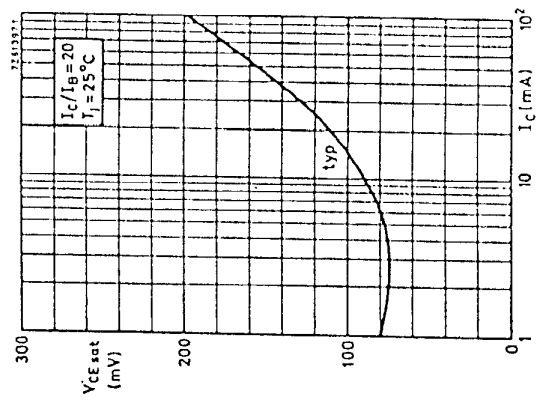
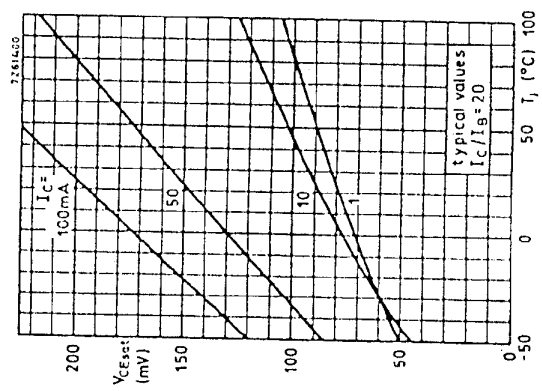
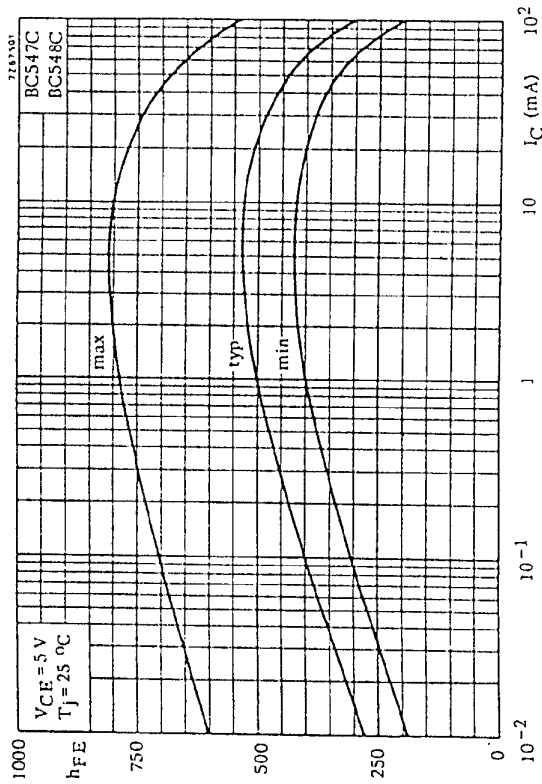
Small signal current gain

$h_{fe} > 125$   
 $\text{typ. } 220$   
 $< 260$

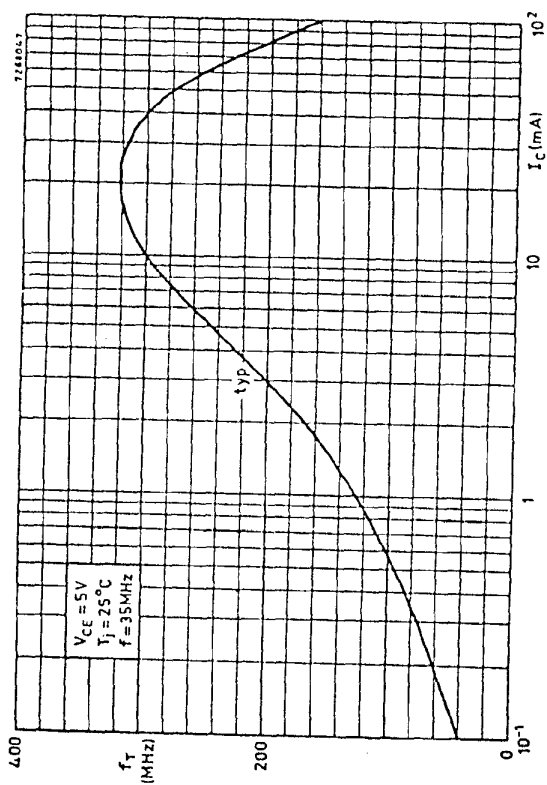
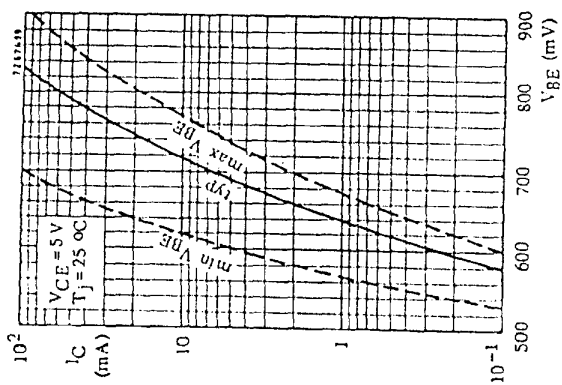
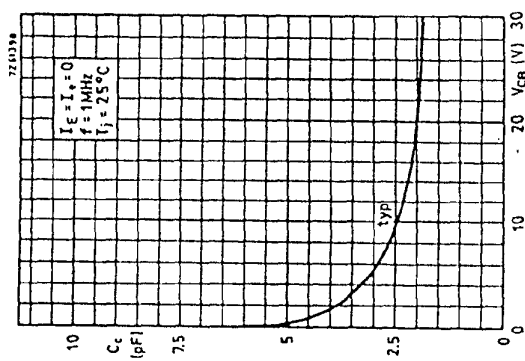
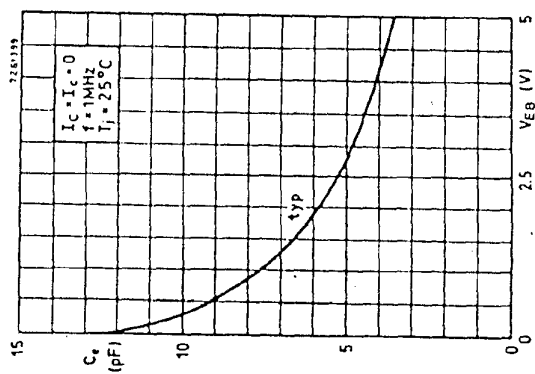
Output admittance

$h_{oc} \text{ typ. } 18$   
 $< 30$

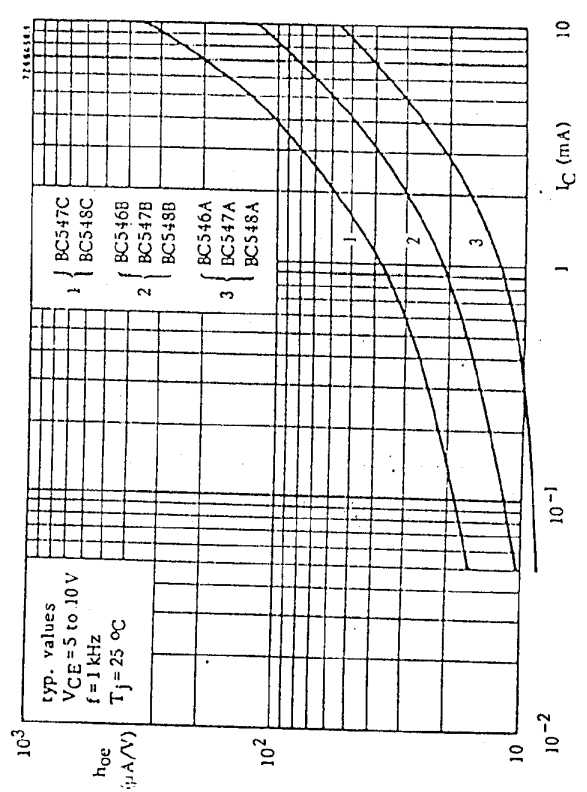
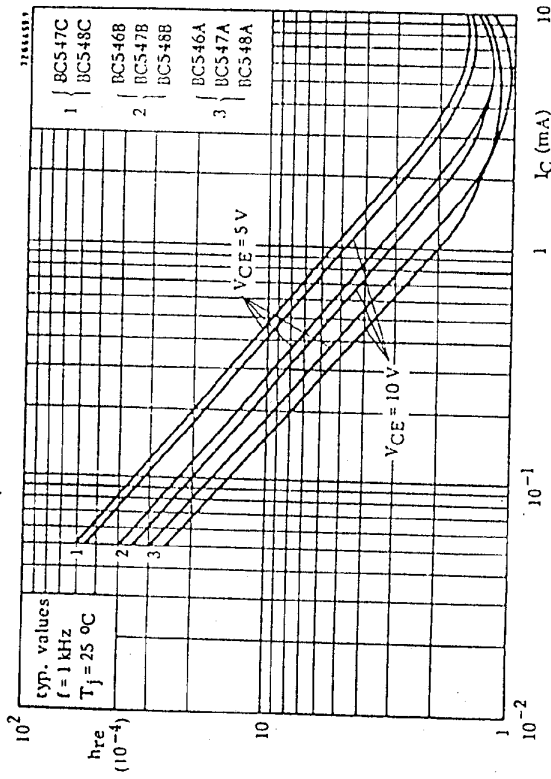
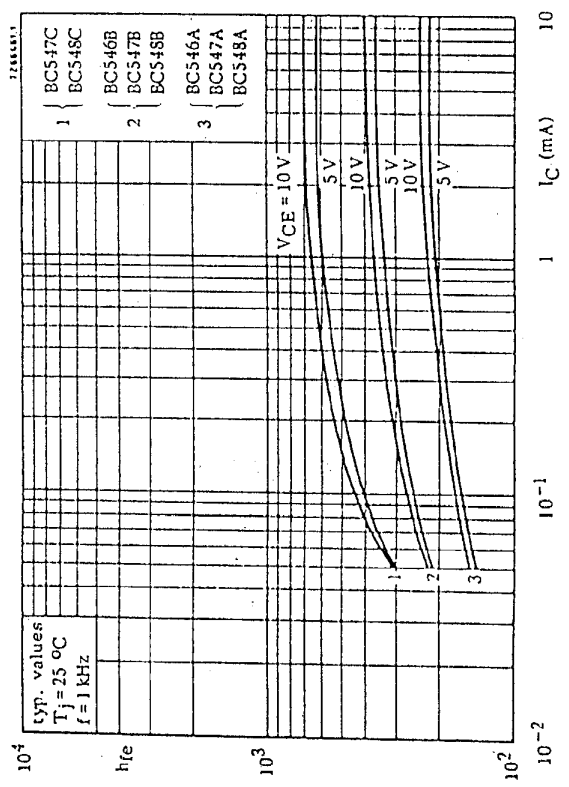
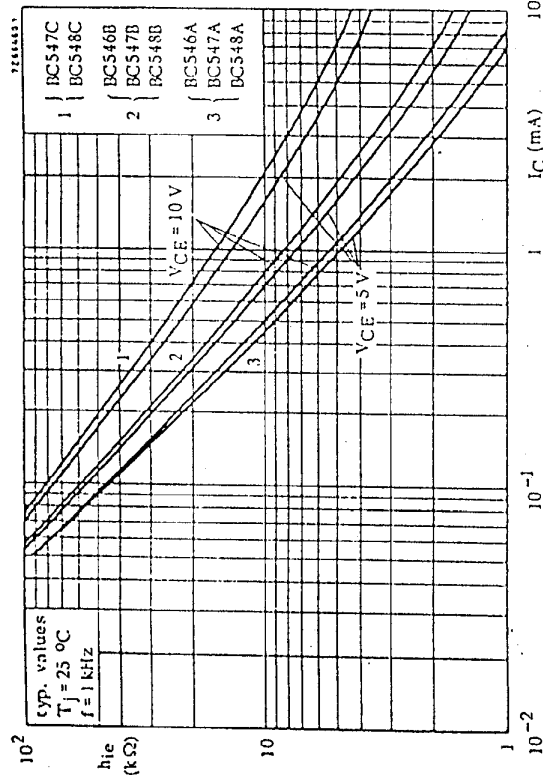
**BC546/548**



**BC546/548**



**BC546/548**



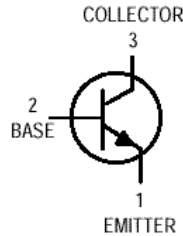
**BC546/548**

# General Purpose Transistors

## NPN Silicon

**2N3903**  
**2N3904\***

\*Motorola Preferred Device



CASE 29-04, STYLE 1  
TO-92 (TO-226AA)

### MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Collector–Emitter Voltage	$V_{CE0}$	40	Vdc
Collector–Base Voltage	$V_{CB0}$	60	Vdc
Emitter–Base Voltage	$V_{EB0}$	6.0	Vdc
Collector Current — Continuous	$I_C$	200	mAdc
Total Device Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	625 5.0	mW mW/ $^\circ\text{C}$
Total Device Dissipation @ $T_C = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	1.5 12	Watts mW/ $^\circ\text{C}$
Operating and Storage Junction Temperature Range	$T_J, T_{stg}$	-55 to +150	$^\circ\text{C}$

### THERMAL CHARACTERISTICS(1)

Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Ambient	$R_{\theta JA}$	200	$^\circ\text{C/W}$
Thermal Resistance, Junction to Case	$R_{\theta JC}$	83.3	$^\circ\text{C/W}$

### ELECTRICAL CHARACTERISTICS ( $T_A = 25^\circ\text{C}$ unless otherwise noted)

Characteristic	Symbol	Min	Max	Unit
----------------	--------	-----	-----	------

### OFF CHARACTERISTICS

Collector–Emitter Breakdown Voltage (2) ( $I_C = 1.0 \text{ mAdc}, I_B = 0$ )	$V_{(BR)CEO}$	40	—	Vdc
Collector–Base Breakdown Voltage ( $I_C = 10 \mu\text{Adc}, I_E = 0$ )	$V_{(BR)CBO}$	60	—	Vdc
Emitter–Base Breakdown Voltage ( $I_E = 10 \mu\text{Adc}, I_C = 0$ )	$V_{(BR)EBO}$	6.0	—	Vdc
Base Cutoff Current ( $V_{CE} = 30 \text{ Vdc}, V_{EB} = 3.0 \text{ Vdc}$ )	$I_{BL}$	—	50	nAdc
Collector Cutoff Current ( $V_{CE} = 30 \text{ Vdc}, V_{EB} = 3.0 \text{ Vdc}$ )	$I_{CEX}$	—	50	nAdc

1. Indicates Data in addition to JEDEC Requirements.
2. Pulse Test: Pulse Width  $\leq 300 \mu\text{s}$ ; Duty Cycle  $\leq 2.0\%$ .



**2N3903/04**

**ELECTRICAL CHARACTERISTICS** ( $T_A = 25^\circ\text{C}$  unless otherwise noted) (Continued)

Characteristic		Symbol	Min	Max	Unit
<b>ON CHARACTERISTICS</b>					
DC Current Gain <sup>(1)</sup> ( $I_C = 0.1 \text{ mAdc}$ , $V_{CE} = 1.0 \text{ Vdc}$ )	2N3903	$h_{FE}$	20	—	—
	2N3904		40	—	
	2N3903		35	—	
	2N3904		70	—	
	2N3903		50	150	
( $I_C = 1.0 \text{ mAdc}$ , $V_{CE} = 1.0 \text{ Vdc}$ )	2N3904	100	300		
( $I_C = 10 \text{ mAdc}$ , $V_{CE} = 1.0 \text{ Vdc}$ )	2N3903	30	—		
2N3904	60	—			
( $I_C = 50 \text{ mAdc}$ , $V_{CE} = 1.0 \text{ Vdc}$ )	2N3903	15	—		
2N3904	30	—			
Collector–Emitter Saturation Voltage <sup>(1)</sup> ( $I_C = 10 \text{ mAdc}$ , $I_B = 1.0 \text{ mAdc}$ ) ( $I_C = 50 \text{ mAdc}$ , $I_B = 5.0 \text{ mAdc}$ )		$V_{CE(sat)}$	—	0.2 0.3	Vdc
Base–Emitter Saturation Voltage <sup>(1)</sup> ( $I_C = 10 \text{ mAdc}$ , $I_B = 1.0 \text{ mAdc}$ ) ( $I_C = 50 \text{ mAdc}$ , $I_B = 5.0 \text{ mAdc}$ )		$V_{BE(sat)}$	0.65 —	0.85 0.95	Vdc

**SMALL–SIGNAL CHARACTERISTICS**

Current–Gain — Bandwidth Product ( $I_C = 10 \text{ mAdc}$ , $V_{CE} = 20 \text{ Vdc}$ , $f = 100 \text{ MHz}$ )	2N3903 2N3904	$f_T$	250 300	— —	MHz
Output Capacitance ( $V_{CB} = 5.0 \text{ Vdc}$ , $I_E = 0$ , $f = 1.0 \text{ MHz}$ )		$C_{obo}$	—	4.0	pF
Input Capacitance ( $V_{EB} = 0.5 \text{ Vdc}$ , $I_C = 0$ , $f = 1.0 \text{ MHz}$ )		$C_{ibo}$	—	8.0	pF
Input Impedance ( $I_C = 1.0 \text{ mAdc}$ , $V_{CE} = 10 \text{ Vdc}$ , $f = 1.0 \text{ kHz}$ )	2N3903 2N3904	$h_{ie}$	1.0 1.0	8.0 10	k $\Omega$
Voltage Feedback Ratio ( $I_C = 1.0 \text{ mAdc}$ , $V_{CE} = 10 \text{ Vdc}$ , $f = 1.0 \text{ kHz}$ )	2N3903 2N3904	$h_{re}$	0.1 0.5	5.0 8.0	$\times 10^{-4}$
Small–Signal Current Gain ( $I_C = 1.0 \text{ mAdc}$ , $V_{CE} = 10 \text{ Vdc}$ , $f = 1.0 \text{ kHz}$ )	2N3903 2N3904	$h_{fe}$	50 100	200 400	—
Output Admittance ( $I_C = 1.0 \text{ mAdc}$ , $V_{CE} = 10 \text{ Vdc}$ , $f = 1.0 \text{ kHz}$ )		$h_{oe}$	1.0	40	$\mu\text{mhos}$
Noise Figure ( $I_C = 100 \mu\text{A}$ , $V_{CE} = 5.0 \text{ Vdc}$ , $R_S = 1.0 \text{ k } \Omega$ , $f = 1.0 \text{ kHz}$ )	2N3903 2N3904	NF	— —	6.0 5.0	dB

**SWITCHING CHARACTERISTICS**

Delay Time	$(V_{CC} = 3.0 \text{ Vdc}$ , $V_{BE} = 0.5 \text{ Vdc}$ , $I_C = 10 \text{ mAdc}$ , $I_{B1} = 1.0 \text{ mAdc}$ )		$t_d$	—	35	ns
Rise Time			$t_r$	—	35	ns
Storage Time	$(V_{CC} = 3.0 \text{ Vdc}$ , $I_C = 10 \text{ mAdc}$ , $I_{B1} = I_{B2} = 1.0 \text{ mAdc}$ )	2N3903 2N3904	$t_s$	—	175 200	ns
Fall Time			$t_f$	—	50	ns

1. Pulse Test: Pulse Width  $\leq 300 \mu\text{s}$ ; Duty Cycle  $\leq 2.0\%$ .

**2N3903/04**

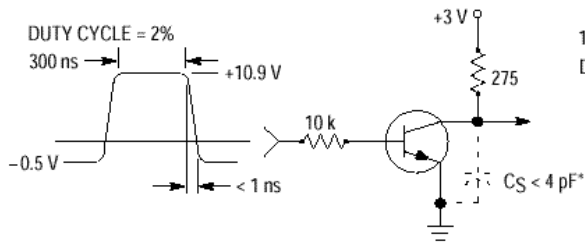


Figure 1. Delay and Rise Time Equivalent Test Circuit

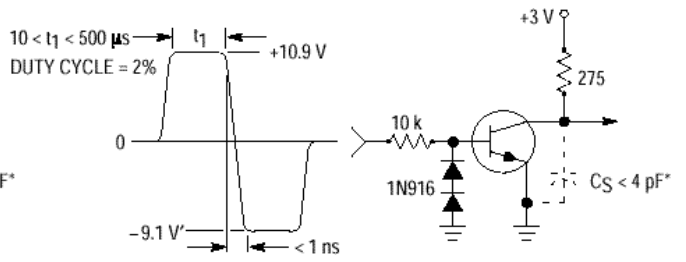


Figure 2. Storage and Fall Time Equivalent Test Circuit

\* Total shunt capacitance of test jig and connectors

TYPICAL TRANSIENT CHARACTERISTICS

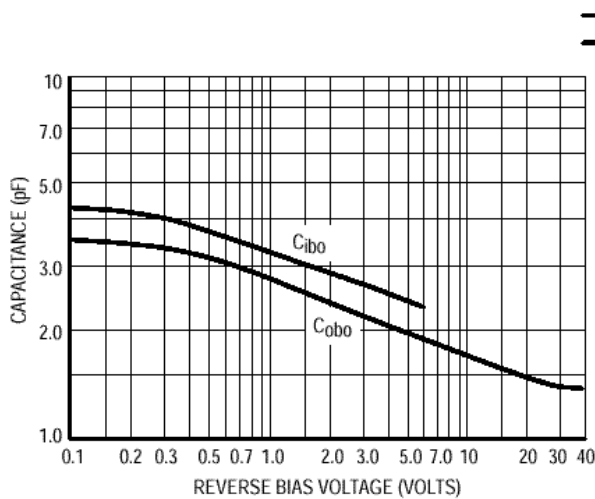


Figure 3. Capacitance

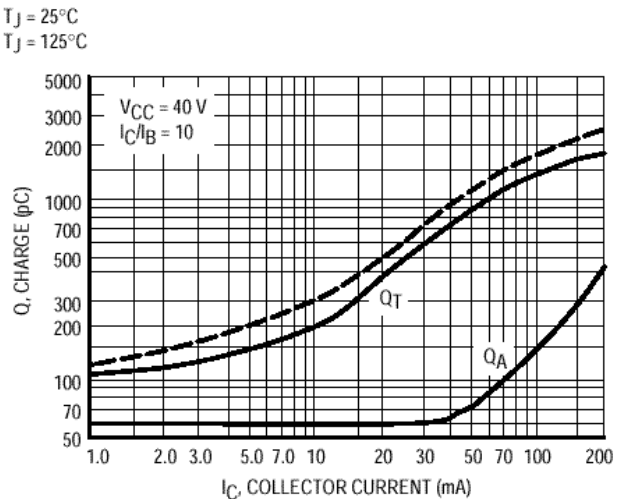


Figure 4. Charge Data

2N3903/04

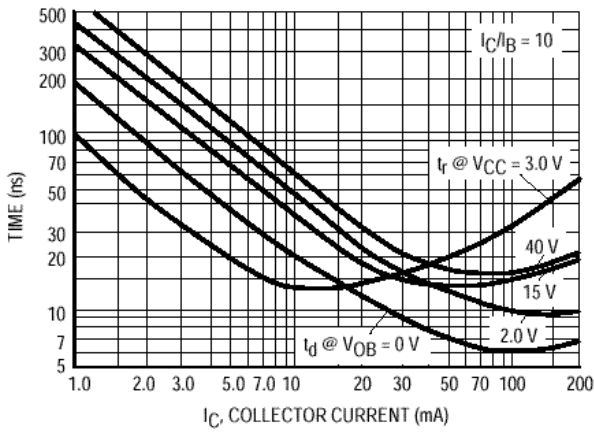


Figure 5. Turn-On Time

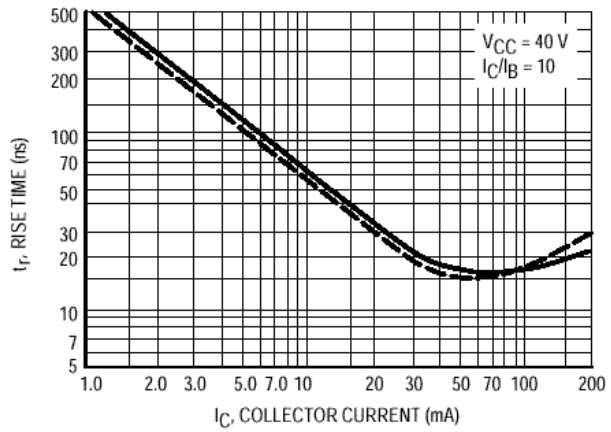


Figure 6. Rise Time

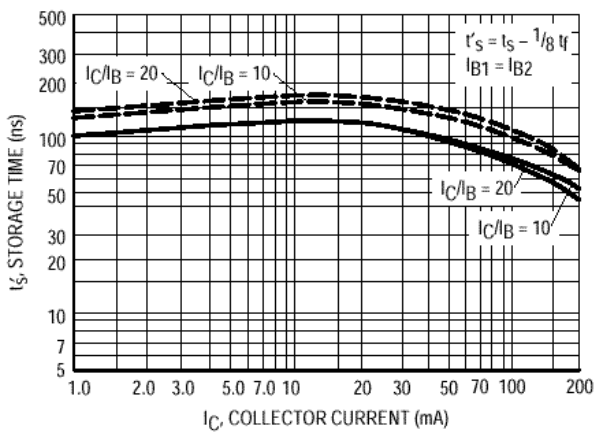


Figure 7. Storage Time

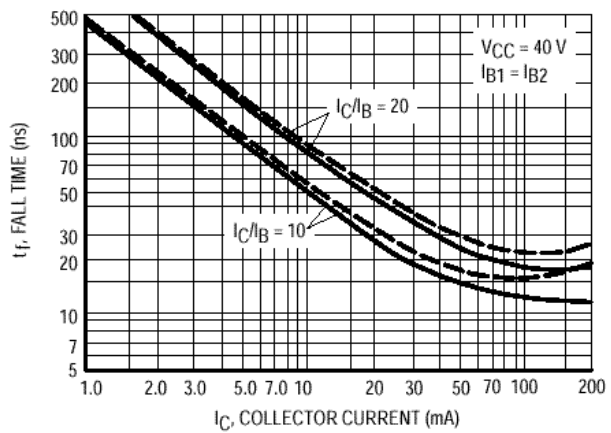


Figure 8. Fall Time

TYPICAL AUDIO SMALL-SIGNAL CHARACTERISTICS  
NOISE FIGURE VARIATIONS

( $V_{CE} = 5.0 \text{ Vdc}$ ,  $T_A = 25^\circ\text{C}$ , Bandwidth = 1.0 Hz)

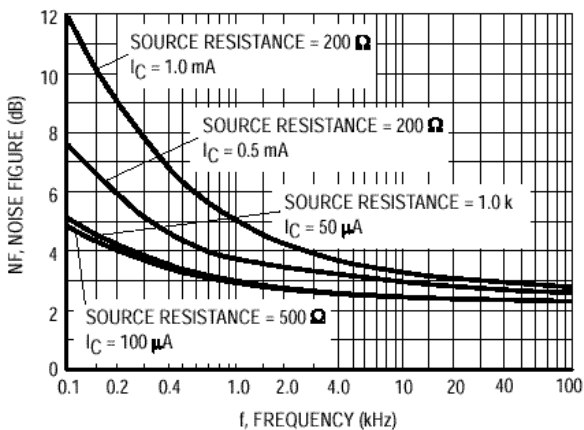


Figure 9.

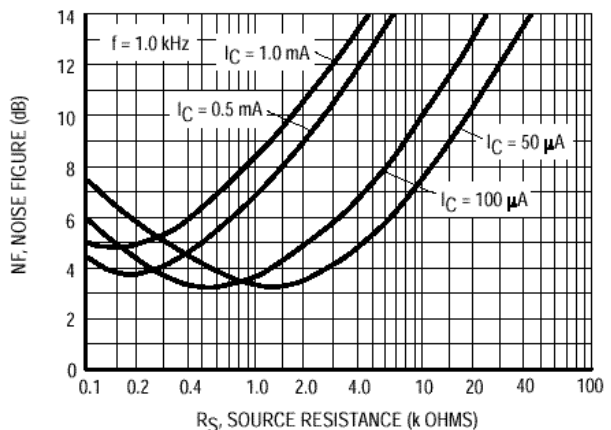


Figure 10.

2N3903/04

**h PARAMETERS**  
 ( $V_{CE} = 10 \text{ Vdc}$ ,  $f = 1.0 \text{ kHz}$ ,  $T_A = 25^\circ\text{C}$ )

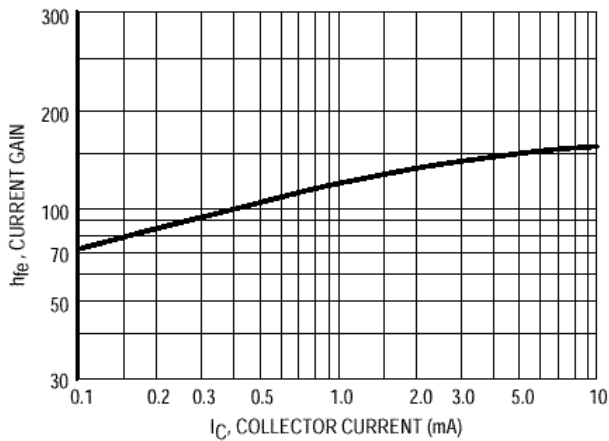


Figure 11. Current Gain

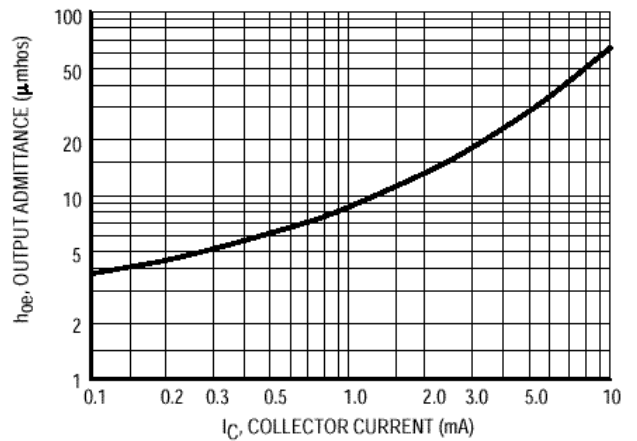


Figure 12. Output Admittance

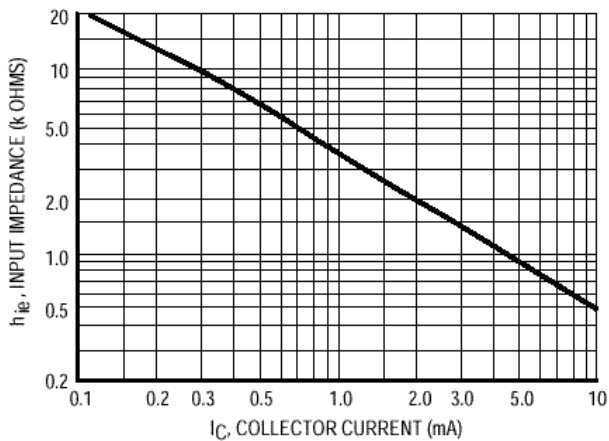


Figure 13. Input Impedance

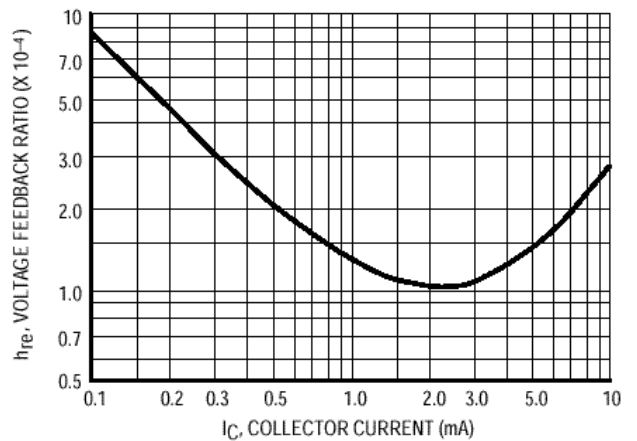


Figure 14. Voltage Feedback Ratio

**TYPICAL STATIC CHARACTERISTICS**

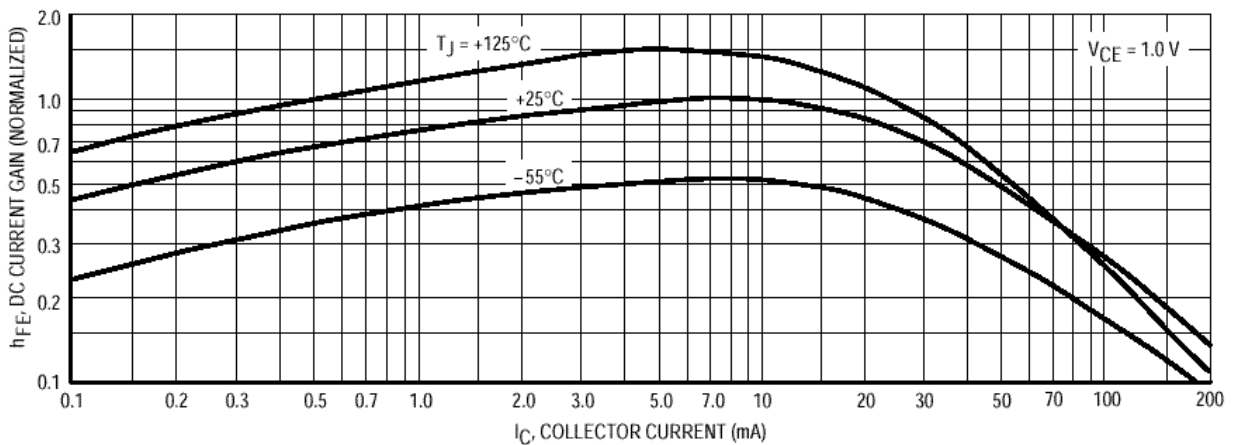


Figure 15. DC Current Gain

**2N3903/04**

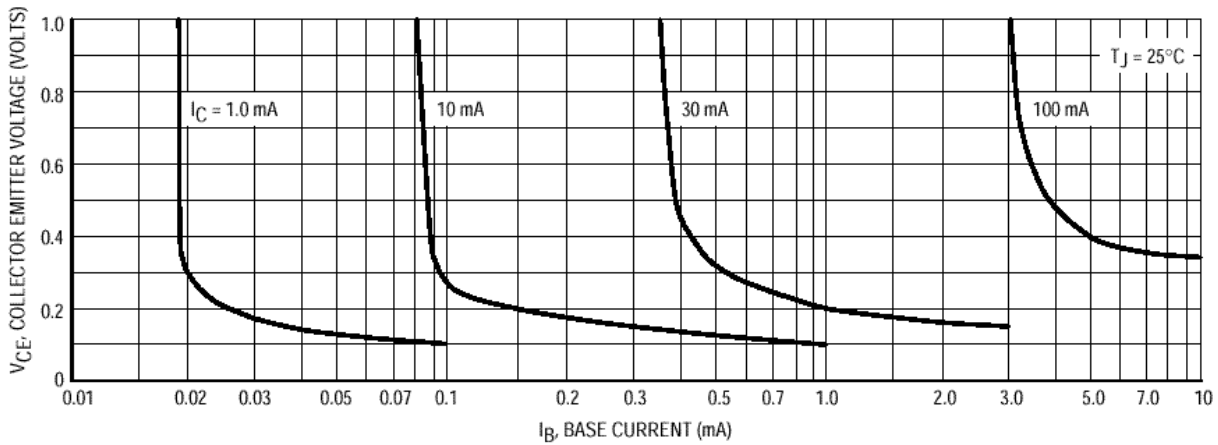


Figure 16. Collector Saturation Region

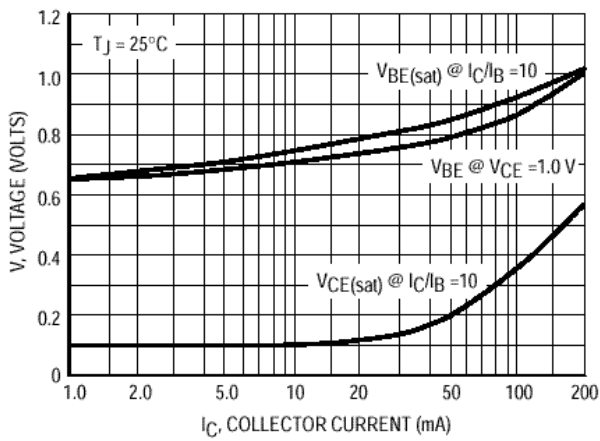


Figure 17. "ON" Voltages

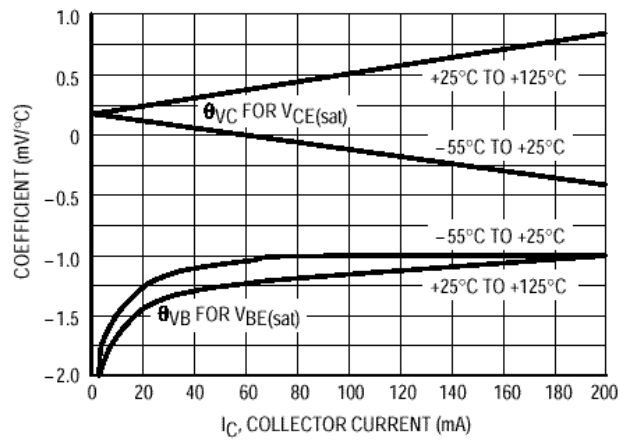
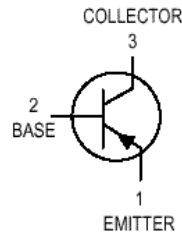


Figure 18. Temperature Coefficients

2N3903/04

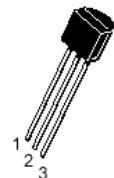
# General Purpose Transistors

## PNP Silicon



**2N3905**  
**2N3906\***

\*Motorola Preferred Device



CASE 29-04, STYLE 1  
TO-92 (TO-226AA)

### MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Collector–Emitter Voltage	$V_{CEO}$	40	Vdc
Collector–Base Voltage	$V_{CB0}$	40	Vdc
Emitter–Base Voltage	$V_{EBO}$	5.0	Vdc
Collector Current — Continuous	$I_C$	200	mAdc
Total Device Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	625 5.0	mW mW/ $^\circ\text{C}$
Total Power Dissipation @ $T_A = 60^\circ\text{C}$	$P_D$	250	mW
Total Device Dissipation @ $T_C = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	1.5 12	Watts mW/ $^\circ\text{C}$
Operating and Storage Junction Temperature Range	$T_J, T_{stg}$	-55 to +150	$^\circ\text{C}$

### THERMAL CHARACTERISTICS(1)

Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Ambient	$R_{\theta JA}$	200	$^\circ\text{C}/\text{W}$
Thermal Resistance, Junction to Case	$R_{\theta JC}$	83.3	$^\circ\text{C}/\text{W}$

### ELECTRICAL CHARACTERISTICS ( $T_A = 25^\circ\text{C}$ unless otherwise noted)

Characteristic	Symbol	Min	Max	Unit
----------------	--------	-----	-----	------

### OFF CHARACTERISTICS

Collector–Emitter Breakdown Voltage (2) ( $I_C = 1.0 \text{ mAdc}, I_B = 0$ )	$V_{(BR)CEO}$	40	—	Vdc
Collector–Base Breakdown Voltage ( $I_C = 10 \mu\text{Adc}, I_E = 0$ )	$V_{(BR)CBO}$	40	—	Vdc
Emitter–Base Breakdown Voltage ( $I_E = 10 \mu\text{Adc}, I_C = 0$ )	$V_{(BR)EBO}$	5.0	—	Vdc
Base Cutoff Current ( $V_{CE} = 30 \text{ Vdc}, V_{EB} = 3.0 \text{ Vdc}$ )	$I_{BL}$	—	50	nAdc
Collector Cutoff Current ( $V_{CE} = 30 \text{ Vdc}, V_{EB} = 3.0 \text{ Vdc}$ )	$I_{CEX}$	—	50	nAdc

1. Indicates Data in addition to JEDEC Requirements.
2. Pulse Test: Pulse Width  $\leq 300 \mu\text{s}$ ; Duty Cycle  $\leq 2.0\%$ .

**2N3905/06**

**ELECTRICAL CHARACTERISTICS** ( $T_A = 25^\circ\text{C}$  unless otherwise noted) (Continued)

Characteristic	Symbol	Min	Max	Unit
<b>ON CHARACTERISTICS(1)</b>				
DC Current Gain ( $I_C = 0.1 \text{ mAdc}$ , $V_{CE} = 1.0 \text{ Vdc}$ )	$h_{FE}$	30	—	—
2N3905		60	—	
2N3906				
( $I_C = 1.0 \text{ mAdc}$ , $V_{CE} = 1.0 \text{ Vdc}$ )		40	—	
2N3905		80	—	
2N3906				
( $I_C = 10 \text{ mAdc}$ , $V_{CE} = 1.0 \text{ Vdc}$ )		50	150	
2N3905		100	300	
2N3906				
( $I_C = 50 \text{ mAdc}$ , $V_{CE} = 1.0 \text{ Vdc}$ )		30	—	
2N3905		60	—	
2N3906				
( $I_C = 100 \text{ mAdc}$ , $V_{CE} = 1.0 \text{ Vdc}$ )		15	—	
2N3905		30	—	
2N3906				
Collector–Emitter Saturation Voltage ( $I_C = 10 \text{ mAdc}$ , $I_B = 1.0 \text{ mAdc}$ ) ( $I_C = 50 \text{ mAdc}$ , $I_B = 5.0 \text{ mAdc}$ )	$V_{CE(sat)}$	—	0.25 0.4	Vdc
Base–Emitter Saturation Voltage ( $I_C = 10 \text{ mAdc}$ , $I_B = 1.0 \text{ mAdc}$ ) ( $I_C = 50 \text{ mAdc}$ , $I_B = 5.0 \text{ mAdc}$ )	$V_{BE(sat)}$	0.65 —	0.85 0.95	Vdc

**SMALL–SIGNAL CHARACTERISTICS**

Current–Gain — Bandwidth Product ( $I_C = 10 \text{ mAdc}$ , $V_{CE} = 20 \text{ Vdc}$ , $f = 100 \text{ MHz}$ )	$f_T$	200 250	— —	MHz
Output Capacitance ( $V_{CB} = 5.0 \text{ Vdc}$ , $I_E = 0$ , $f = 1.0 \text{ MHz}$ )	$C_{obo}$	—	4.5	pF
Input Capacitance ( $V_{EB} = 0.5 \text{ Vdc}$ , $I_C = 0$ , $f = 1.0 \text{ MHz}$ )	$C_{ibo}$	—	10.0	pF
Input Impedance ( $I_C = 1.0 \text{ mAdc}$ , $V_{CE} = 10 \text{ Vdc}$ , $f = 1.0 \text{ kHz}$ )	$h_{ie}$	0.5 2.0	8.0 12	k $\Omega$
Voltage Feedback Ratio ( $I_C = 1.0 \text{ mAdc}$ , $V_{CE} = 10 \text{ Vdc}$ , $f = 1.0 \text{ kHz}$ )	$h_{re}$	0.1 0.1	5.0 10	$\times 10^{-4}$
Small–Signal Current Gain ( $I_C = 1.0 \text{ mAdc}$ , $V_{CE} = 10 \text{ Vdc}$ , $f = 1.0 \text{ kHz}$ )	$h_{fe}$	50 100	200 400	—
Output Admittance ( $I_C = 1.0 \text{ mAdc}$ , $V_{CE} = 10 \text{ Vdc}$ , $f = 1.0 \text{ kHz}$ )	$h_{oe}$	1.0 3.0	40 60	$\mu\text{mhos}$
Noise Figure ( $I_C = 100 \mu\text{Adc}$ , $V_{CE} = 5.0 \text{ Vdc}$ , $R_S = 1.0 \text{ k}\Omega$ , $f = 1.0 \text{ kHz}$ )	NF	— —	5.0 4.0	dB

**SWITCHING CHARACTERISTICS**

Delay Time	$(V_{CC} = 3.0 \text{ Vdc}$ , $V_{BE} = 0.5 \text{ Vdc}$ , $I_C = 10 \text{ mAdc}$ , $I_{B1} = 1.0 \text{ mAdc}$ )	$t_d$	—	35	ns
Rise Time		$t_r$	—	35	ns
Storage Time	$(V_{CC} = 3.0 \text{ Vdc}$ , $I_C = 10 \text{ mAdc}$ , $I_{B1} = I_{B2} = 1.0 \text{ mAd}$ )	$t_s$	—	200 225	ns
Fall Time		$t_f$	—	60 75	ns

1. Pulse Test: Pulse Width  $\leq 300 \mu\text{s}$ ; Duty Cycle  $\leq 2.0\%$ .

**2N3905/06**



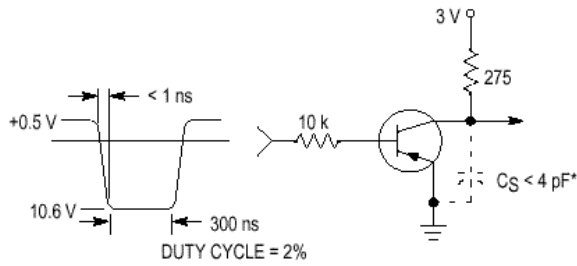


Figure 1. Delay and Rise Time Equivalent Test Circuit

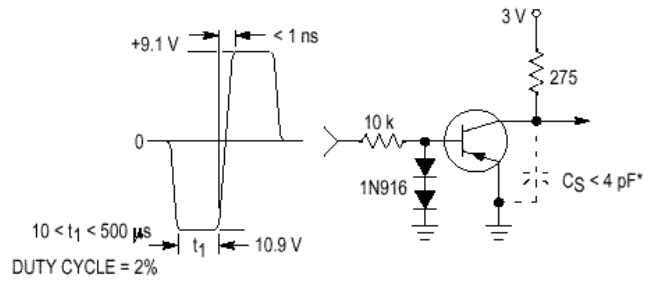


Figure 2. Storage and Fall Time Equivalent Test Circuit

\* Total shunt capacitance of test jig and connectors

TYPICAL TRANSIENT CHARACTERISTICS

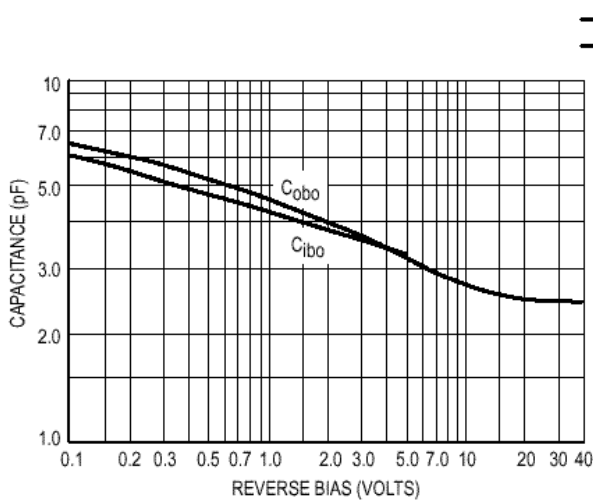


Figure 3. Capacitance

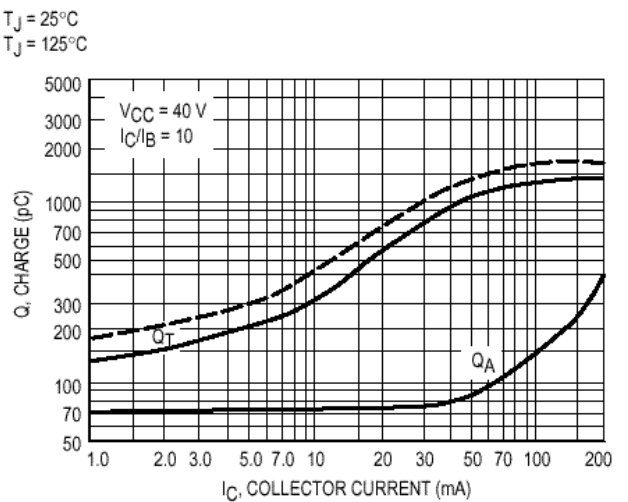


Figure 4. Charge Data

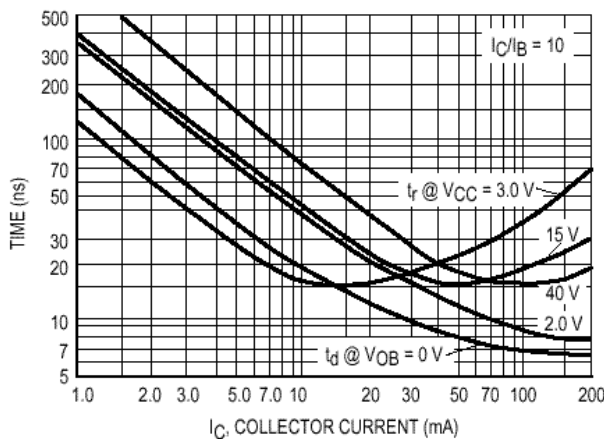


Figure 5. Turn-On Time

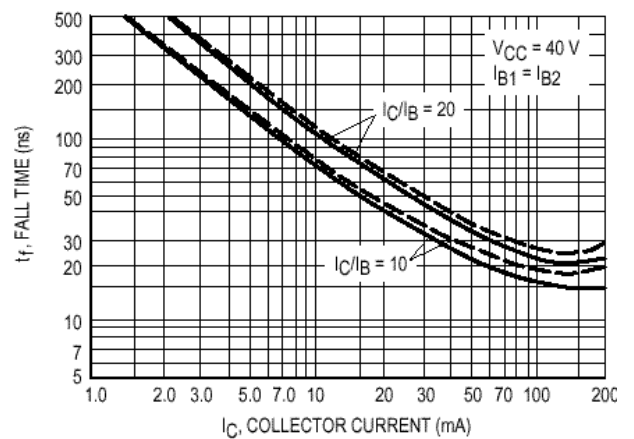


Figure 6. Fall Time

2N3905/06

**TYPICAL AUDIO SMALL-SIGNAL CHARACTERISTICS  
NOISE FIGURE VARIATIONS**

( $V_{CE} = -5.0$  Vdc,  $T_A = 25^\circ\text{C}$ , Bandwidth = 1.0 Hz)

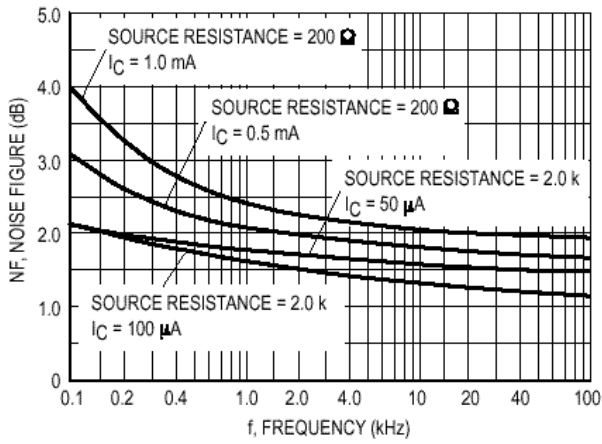


Figure 7.

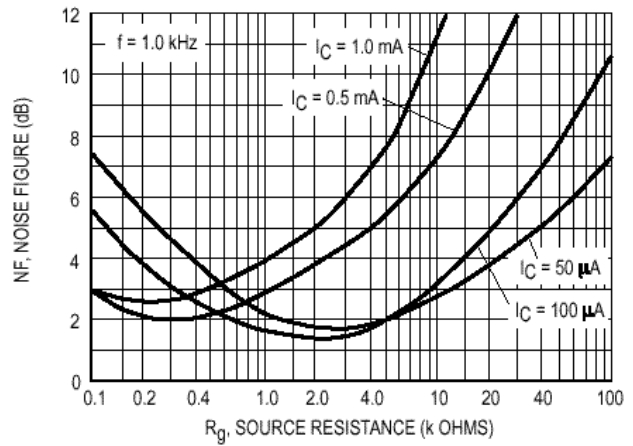


Figure 8.

**h PARAMETERS**

( $V_{CE} = -10$  Vdc,  $f = 1.0$  kHz,  $T_A = 25^\circ\text{C}$ )

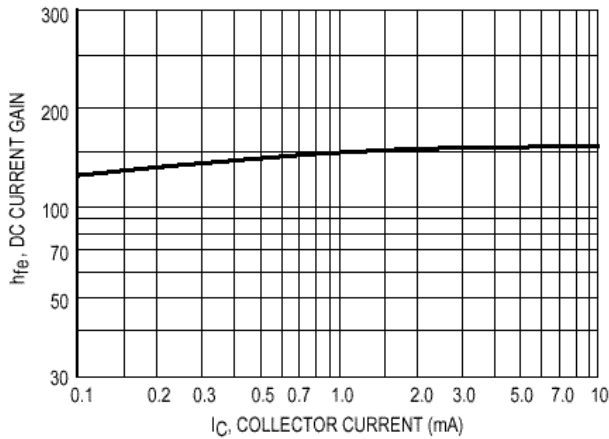


Figure 9. Current Gain

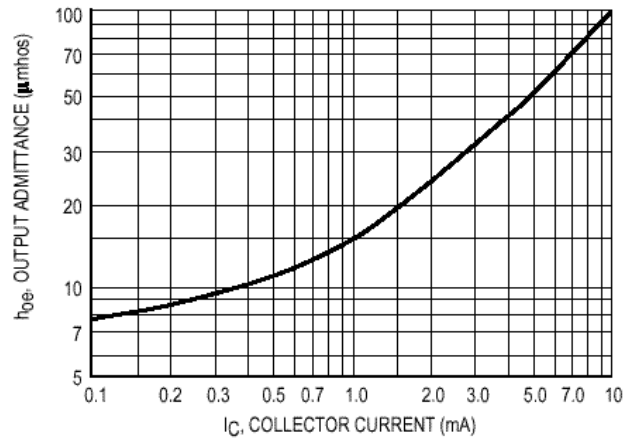


Figure 10. Output Admittance

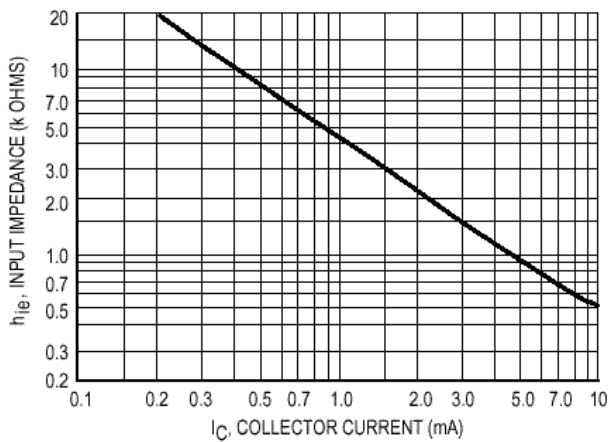


Figure 11. Input Impedance

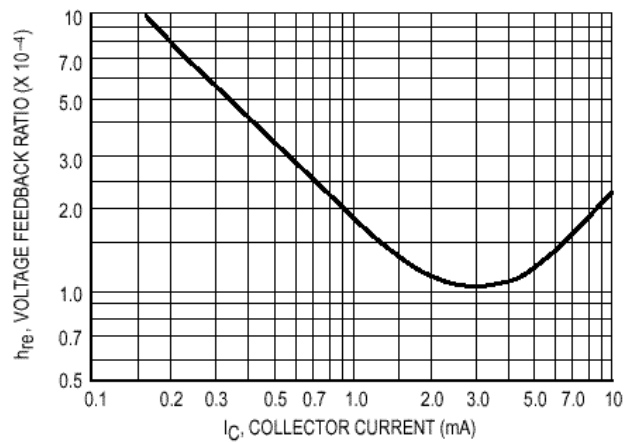


Figure 12. Voltage Feedback Ratio

**2N3905/06**

TYPICAL STATIC CHARACTERISTICS

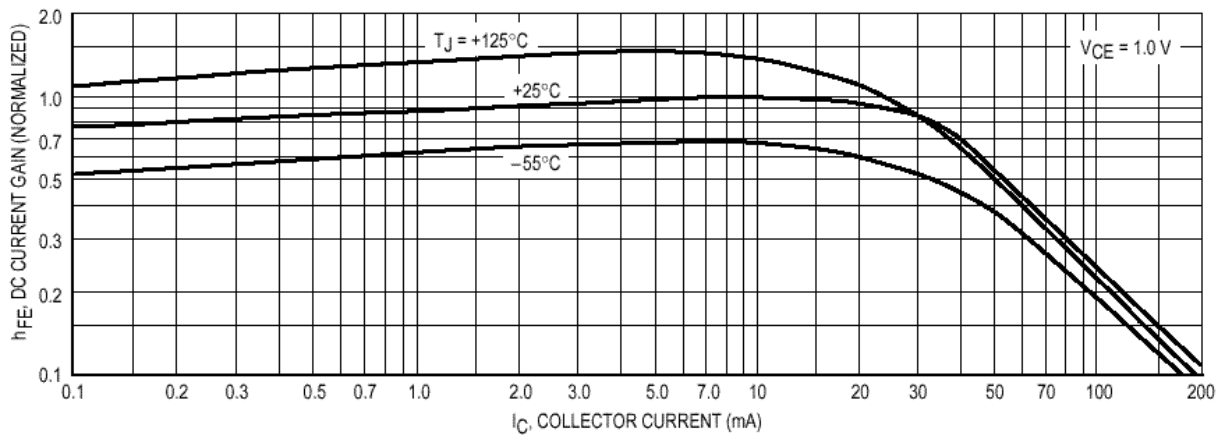


Figure 13. DC Current Gain

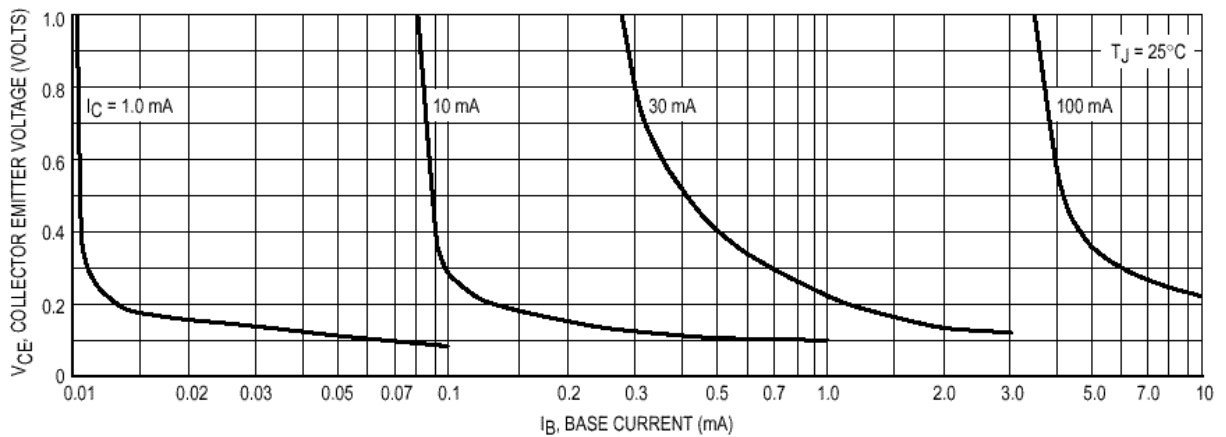


Figure 14. Collector Saturation Region

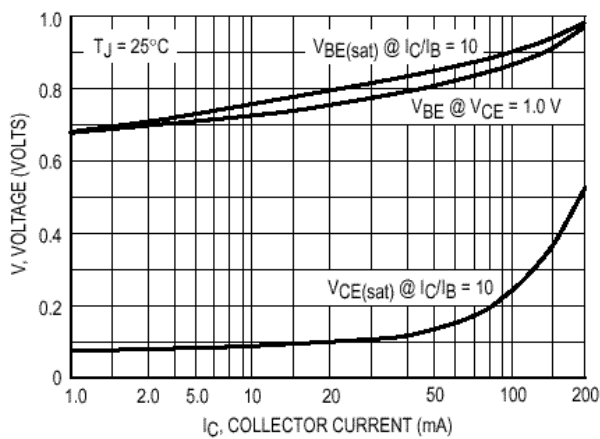


Figure 15. "ON" Voltages

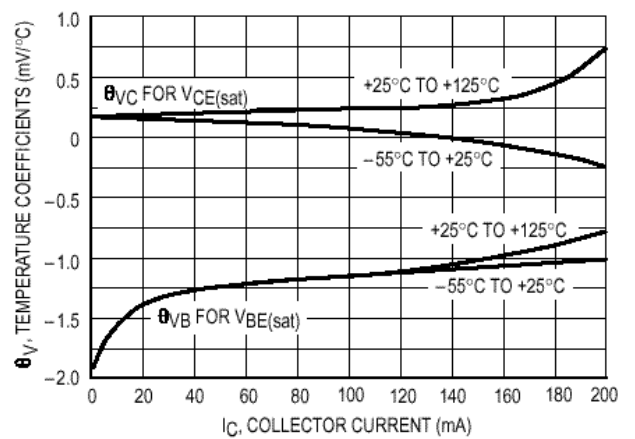
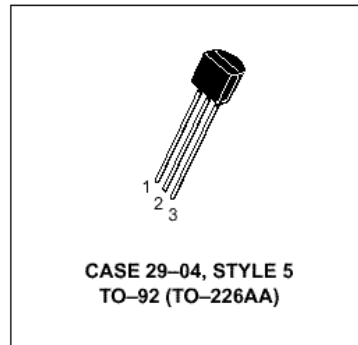
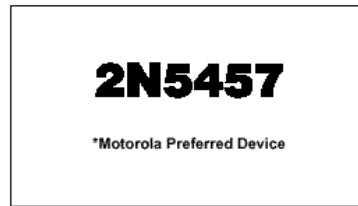
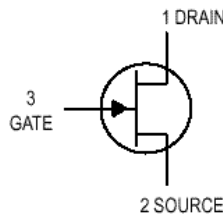


Figure 16. Temperature Coefficients

2N3905/06

# JFETs — General Purpose

## N-Channel — Depletion



### MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Drain-Source Voltage	$V_{DS}$	25	Vdc
Drain-Gate Voltage	$V_{DG}$	25	Vdc
Reverse Gate-Source Voltage	$V_{GSR}$	-25	Vdc
Gate Current	$I_G$	10	mAdc
Total Device Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	310 2.82	mW mW/ $^\circ\text{C}$
Junction Temperature Range	$T_J$	125	$^\circ\text{C}$
Storage Channel Temperature Range	$T_{stg}$	-65 to +150	$^\circ\text{C}$

### ELECTRICAL CHARACTERISTICS ( $T_A = 25^\circ\text{C}$ unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
----------------	--------	-----	-----	-----	------

### OFF CHARACTERISTICS

Gate-Source Breakdown Voltage ( $I_G = -10 \mu\text{Adc}$ , $V_{DS} = 0$ )	$V_{(BR)GSS}$	-25	—	—	Vdc
Gate Reverse Current ( $V_{GS} = -15 \text{Vdc}$ , $V_{DS} = 0$ ) ( $V_{GS} = -15 \text{Vdc}$ , $V_{DS} = 0$ , $T_A = 100^\circ\text{C}$ )	$I_{GSS}$	—	—	-1.0 -200	nAdc
Gate-Source Cutoff Voltage ( $V_{DS} = 15 \text{Vdc}$ , $I_D = 10 \text{nAdc}$ )	$V_{GS(off)}$	-0.5	—	-6.0	Vdc
Gate-Source Voltage ( $V_{DS} = 15 \text{Vdc}$ , $I_D = 100 \mu\text{Adc}$ )	$V_{GS}$	—	-2.5	—	Vdc

### ON CHARACTERISTICS

Zero-Gate-Voltage Drain Current (1) ( $V_{DS} = 15 \text{Vdc}$ , $V_{GS} = 0$ )	$I_{DSS}$	1.0	3.0	5.0	mAdc
--	-----------	-----	-----	-----	------

### SMALL-SIGNAL CHARACTERISTICS

Forward Transfer Admittance Common Source (1) ( $V_{DS} = 15 \text{Vdc}$ , $V_{GS} = 0$ , $f = 1.0 \text{kHz}$ )	$ y_{fs} $	1000	—	5000	$\mu\text{mhos}$
Output Admittance Common Source (1) ( $V_{DS} = 15 \text{Vdc}$ , $V_{GS} = 0$ , $f = 1.0 \text{kHz}$ )	$ y_{os} $	—	10	50	$\mu\text{mhos}$
Input Capacitance ( $V_{DS} = 15 \text{Vdc}$ , $V_{GS} = 0$ , $f = 1.0 \text{MHz}$ )	$C_{iss}$	—	4.5	7.0	pF
Reverse Transfer Capacitance ( $V_{DS} = 15 \text{Vdc}$ , $V_{GS} = 0$ , $f = 1.0 \text{MHz}$ )	$C_{rss}$	—	1.5	3.0	pF

1. Pulse Test; Pulse Width  $\leq 630 \text{ms}$ , Duty Cycle  $\leq 10\%$ .

**2N5457**

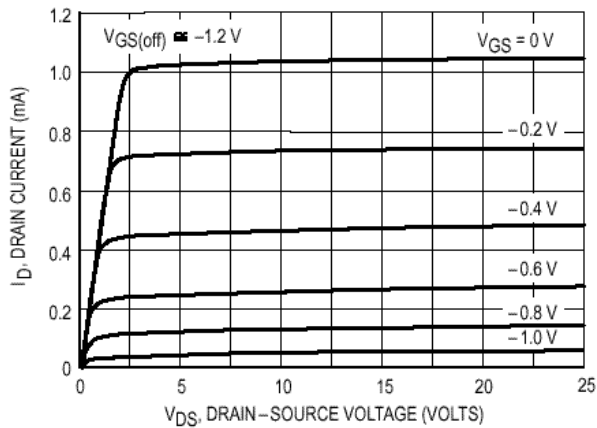


Figure 3. Typical Drain Characteristics

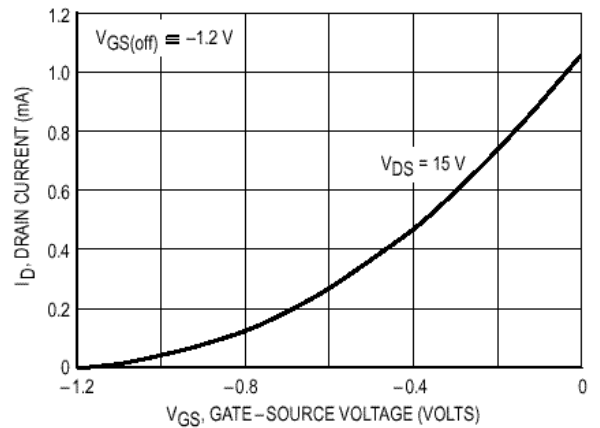


Figure 4. Common Source Transfer Characteristics

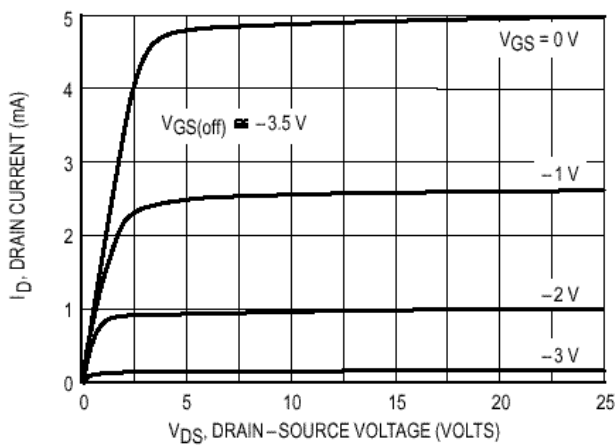


Figure 5. Typical Drain Characteristics

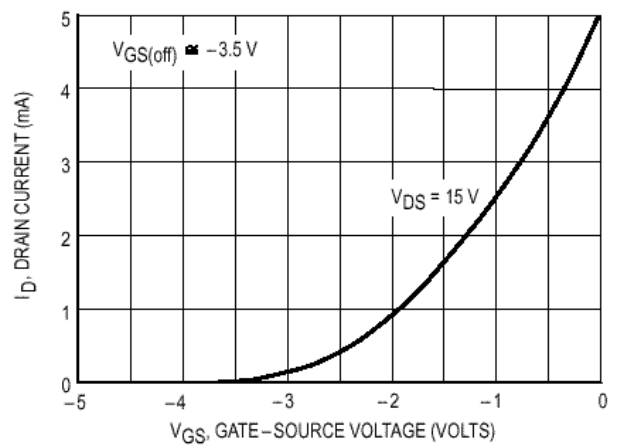


Figure 6. Common Source Transfer Characteristics

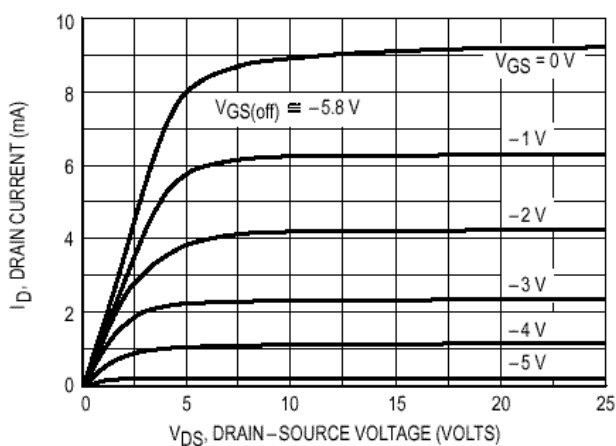


Figure 7. Typical Drain Characteristics

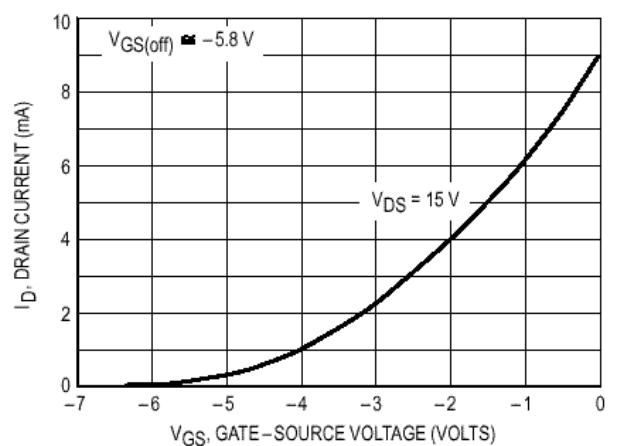
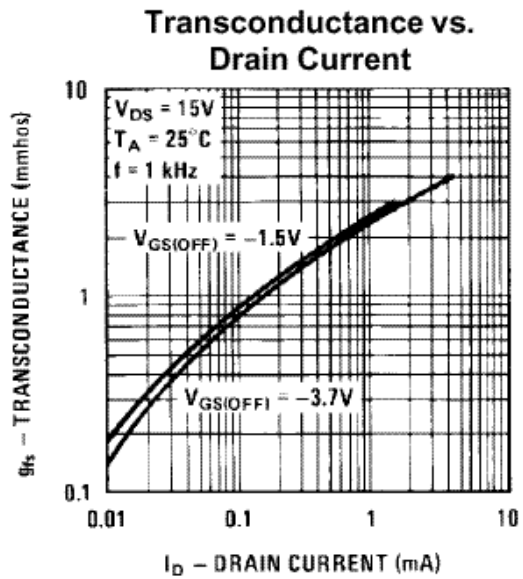
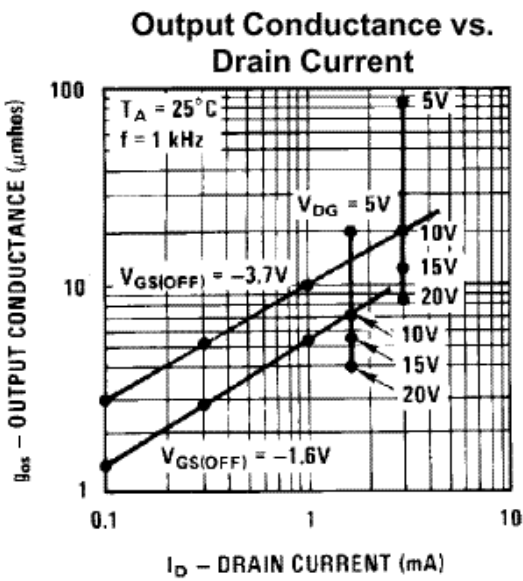
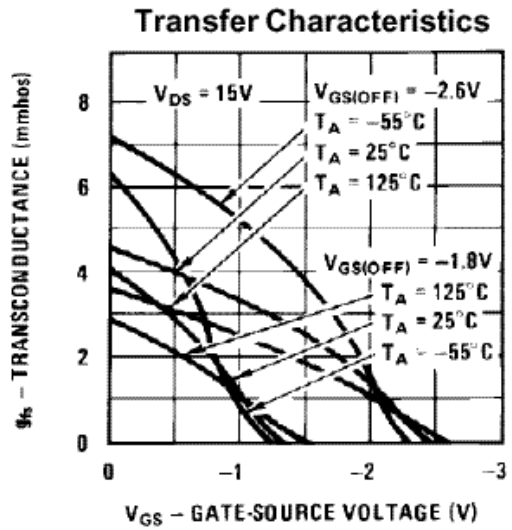
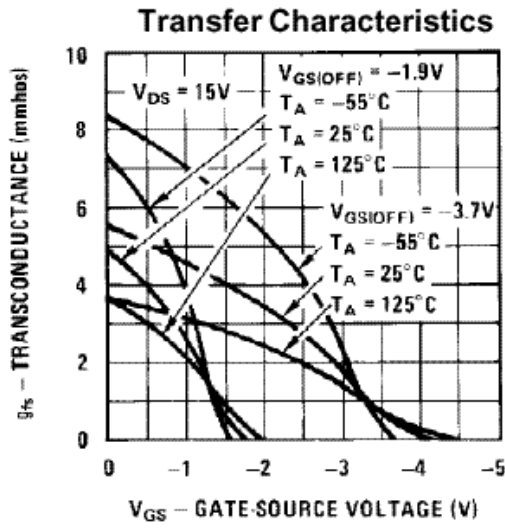


Figure 8. Common Source Transfer Characteristics

**2N5457**

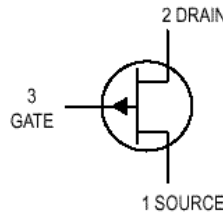
Typical Characteristics (continued)



**2N5457**

# JFET Amplifiers

## P-Channel — Depletion



**2N5460  
thru  
2N5462**



**CASE 29-04, STYLE 7  
TO-92 (TO-226AA)**

### MAXIMUM RATINGS

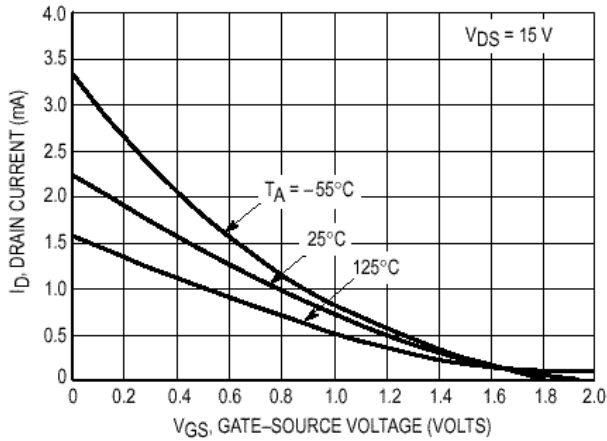
Rating	Symbol	Value	Unit
Drain-Source Voltage	$V_{DS}$	40	Vdc
Reverse Gate-Source Voltage	$V_{GSR}$	40	Vdc
Forward Gate Current	$I_{G(f)}$	10	mAdc
Total Device Dissipation @ $T_A = 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	350 2.8	mW mW/ $^\circ\text{C}$
Junction Temperature Range	$T_J$	-65 to +135	$^\circ\text{C}$
Storage Channel Temperature Range	$T_{stg}$	-65 to +150	$^\circ\text{C}$

### ELECTRICAL CHARACTERISTICS ( $T_A = 25^\circ\text{C}$ unless otherwise noted)

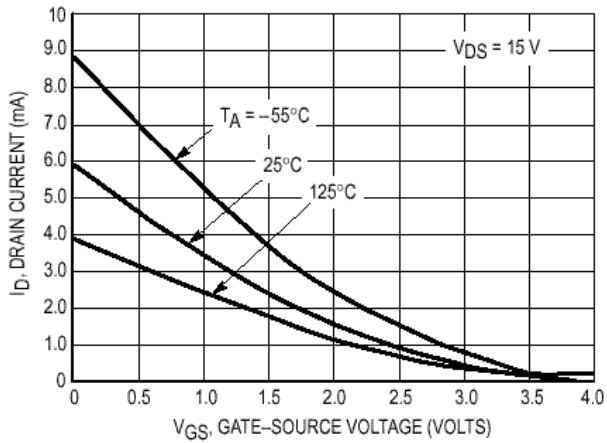
Characteristic	Symbol	Min	Typ	Max	Unit
<b>OFF CHARACTERISTICS</b>					
Gate-Source Breakdown Voltage ( $I_G = 10 \mu\text{Adc}$ , $V_{DS} = 0$ )	2N5460, 2N5461, 2N5462 $V_{(BR)GSS}$	40	—	—	Vdc
Gate Reverse Current ( $V_{GS} = 20 \text{ Vdc}$ , $V_{DS} = 0$ )	2N5460, 2N5461, 2N5462 $I_{GSS}$	—	—	5.0	nAdc
( $V_{GS} = 30 \text{ Vdc}$ , $V_{DS} = 0$ )		—	—	1.0	$\mu\text{Adc}$
( $V_{GS} = 20 \text{ Vdc}$ , $V_{DS} = 0$ , $T_A = 100^\circ\text{C}$ )	2N5460, 2N5461, 2N5462	—	—	1.0	$\mu\text{Adc}$
( $V_{GS} = 30 \text{ Vdc}$ , $V_{DS} = 0$ , $T_A = 100^\circ\text{C}$ )		—	—	1.0	$\mu\text{Adc}$
Gate-Source Cutoff Voltage ( $V_{DS} = 15 \text{ Vdc}$ , $I_D = 1.0 \mu\text{Adc}$ )	2N5460 2N5461 2N5462 $V_{GS(off)}$	0.75 1.0 1.8	— — —	6.0 7.5 9.0	Vdc
Gate-Source Voltage ( $V_{DS} = 15 \text{ Vdc}$ , $I_D = 0.1 \text{ mAdc}$ )	2N5460 $V_{GS}$	0.5	—	4.0	Vdc
( $V_{DS} = 15 \text{ Vdc}$ , $I_D = 0.2 \text{ mAdc}$ )	2N5461	0.8	—	4.5	Vdc
( $V_{DS} = 15 \text{ Vdc}$ , $I_D = 0.4 \text{ mAdc}$ )	2N5462	1.5	—	6.0	Vdc
<b>ON CHARACTERISTICS</b>					
Zero-Gate-Voltage Drain Current ( $V_{DS} = 15 \text{ Vdc}$ , $V_{GS} = 0$ , $f = 1.0 \text{ kHz}$ )	2N5460 2N5461 2N5462 $I_{DSS}$	-1.0 -2.0 -4.0	— — —	-5.0 -9.0 -16	mAdc
<b>SMALL-SIGNAL CHARACTERISTICS</b>					
Forward Transfer Admittance ( $V_{DS} = 15 \text{ Vdc}$ , $V_{GS} = 0$ , $f = 1.0 \text{ kHz}$ )	2N5460 2N5461 2N5462 $ y_{fs} $	1000 1500 2000	— — —	4000 5000 6000	$\mu\text{mhos}$
Output Admittance ( $V_{DS} = 15 \text{ Vdc}$ , $V_{GS} = 0$ , $f = 1.0 \text{ kHz}$ )	$ y_{os} $	—	—	75	$\mu\text{mhos}$
Input Capacitance ( $V_{DS} = 15 \text{ Vdc}$ , $V_{GS} = 0$ , $f = 1.0 \text{ MHz}$ )	$C_{iss}$	—	5.0	7.0	pF
Reverse Transfer Capacitance ( $V_{DS} = 15 \text{ Vdc}$ , $V_{GS} = 0$ , $f = 1.0 \text{ MHz}$ )	$C_{rss}$	—	1.0	2.0	pF
<b>FUNCTIONAL CHARACTERISTICS</b>					
Noise Figure ( $V_{DS} = 15 \text{ Vdc}$ , $V_{GS} = 0$ , $R_G = 1.0 \text{ Megohm}$ , $f = 100 \text{ Hz}$ , $BW = 1.0 \text{ Hz}$ )	NF	—	1.0	2.5	dB
Equivalent Short-Circuit Input Noise Voltage ( $V_{DS} = 15 \text{ Vdc}$ , $V_{GS} = 0$ , $f = 100 \text{ Hz}$ , $BW = 1.0 \text{ Hz}$ )	$e_n$	—	60	115	$\text{nV}/\sqrt{\text{Hz}}$

## 2N5460/62

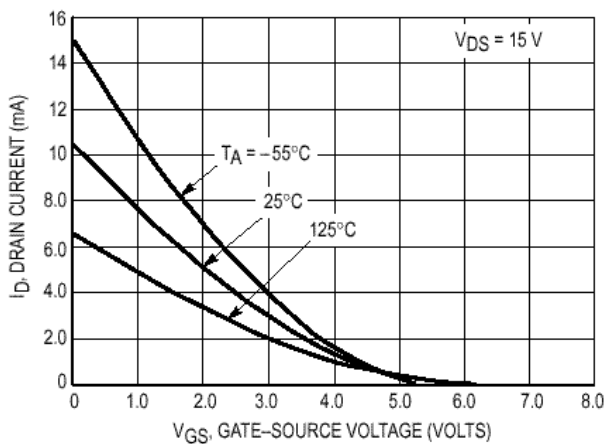
**DRAIN CURRENT versus GATE SOURCE VOLTAGE**



**Figure 1. VGS(off) = 2.0 Volts**

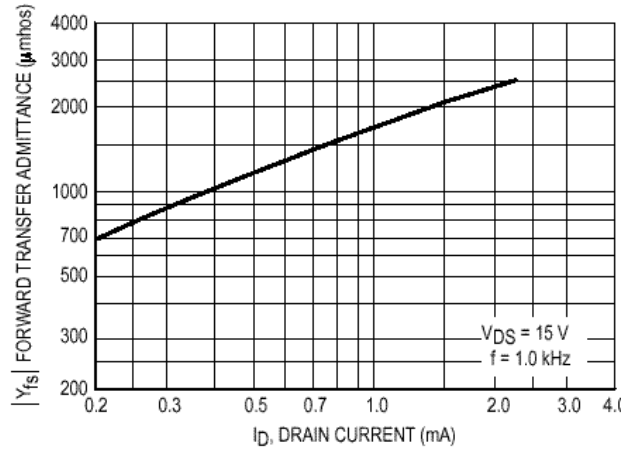


**Figure 2. VGS(off) = 4.0 Volts**

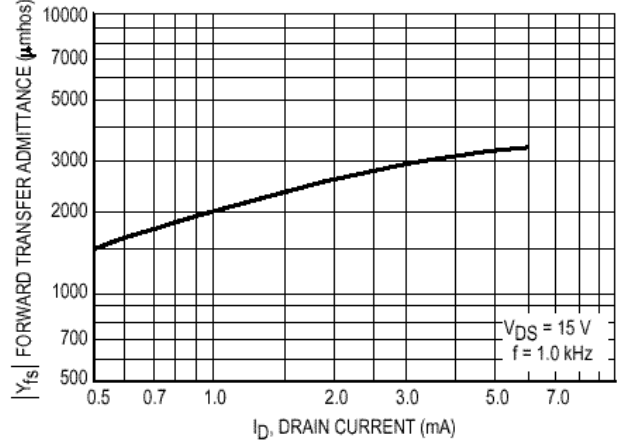


**Figure 3. VGS(off) = 5.0 Volts**

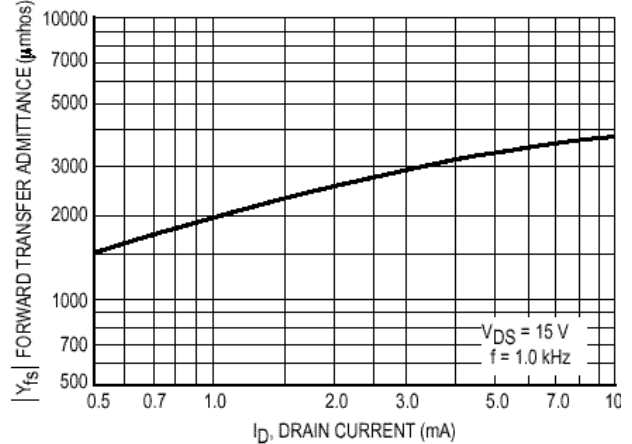
**FORWARD TRANSFER ADMITTANCE versus DRAIN CURRENT**



**Figure 4. VGS(off) = 2.0 Volts**



**Figure 5. VGS(off) = 4.0 Volts**



**Figure 6. VGS(off) = 5.0 Volts**

**2N5460/62**



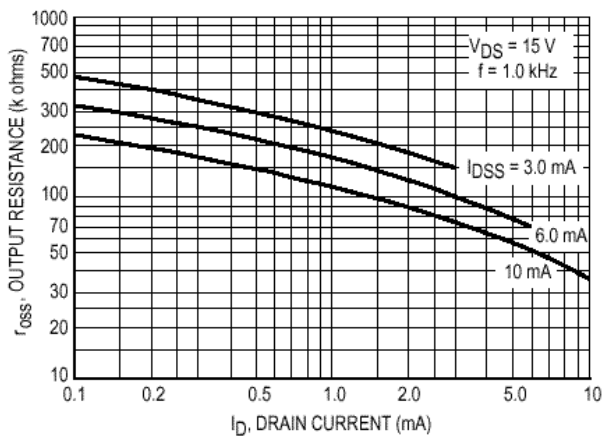


Figure 7. Output Resistance versus Drain Current

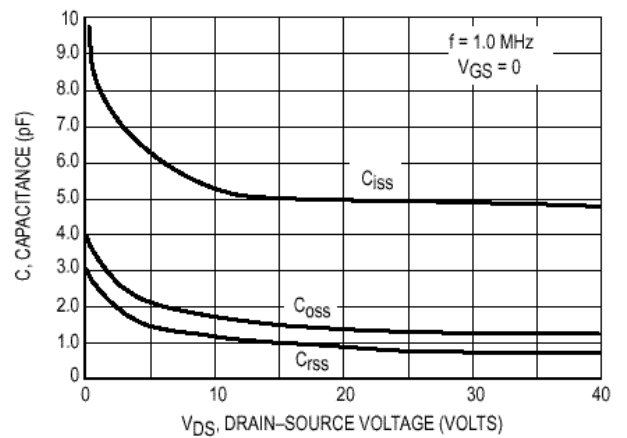


Figure 8. Capacitance versus Drain-Source Voltage

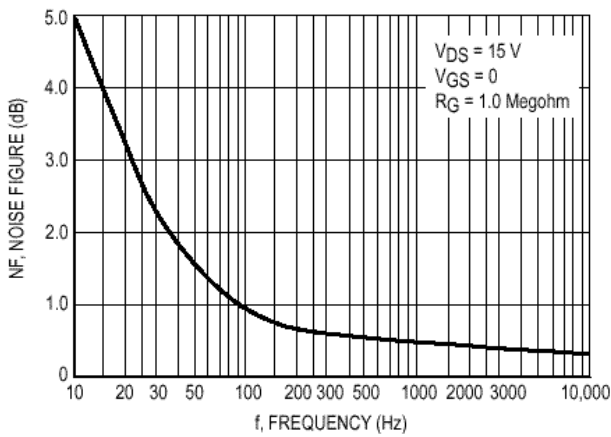


Figure 9. Noise Figure versus Frequency

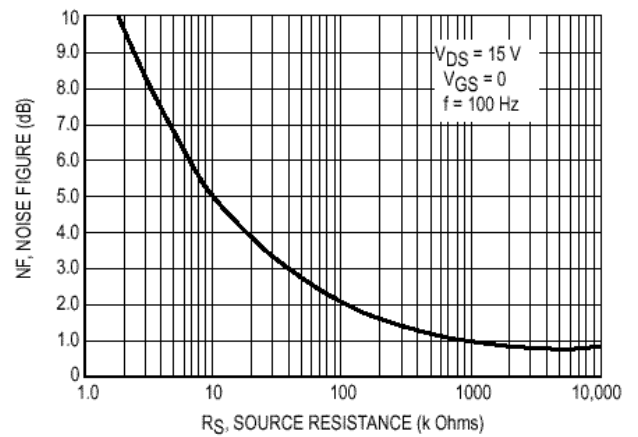
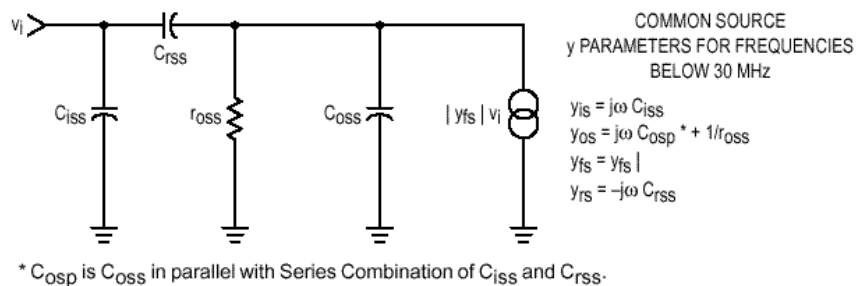


Figure 10. Noise Figure versus Source Resistance



2N5460/62

## General purpose operational amplifier

• A741/• A741C/SA741C

### DESCRIPTION

The • A741 is a high performance operational amplifier with high open-loop gain, internal compensation, high common mode range and exceptional temperature stability. The • A741 is short-circuit-protected and allows for nulling of offset voltage.

### FEATURES

- Internal frequency compensation
- Short circuit protection
- Excellent temperature stability
- High input voltage range

### PIN CONFIGURATION

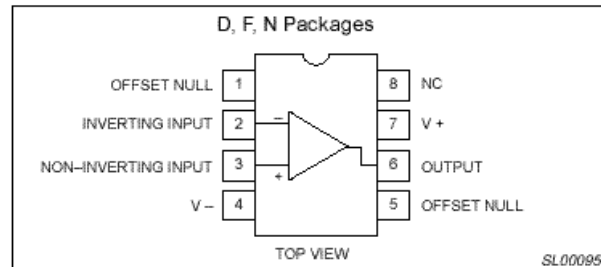


Figure 1. Pin Configuration

### ORDERING INFORMATION

DESCRIPTION	TEMPERATURE RANGE	ORDER CODE	DWG #
8-Pin Plastic Dual In-Line Package (DIP)	-55°C to +125°C	• A741N	SOT97-1
8-Pin Plastic Dual In-Line Package (DIP)	0 to +70°C	• A741CN	SOT97-1
8-Pin Plastic Dual In-Line Package (DIP)	-40°C to +85°C	SA741CN	SOT97-1
8-Pin Ceramic Dual In-Line Package (CERDIP)	-55°C to +125°C	• A741F	0580A
8-Pin Ceramic Dual In-Line Package (CERDIP)	0 to +70°C	• A741CF	0580A
8-Pin Small Outline (SO) Package	0 to +70°C	• A741CD	SOT96-1

### ABSOLUTE MAXIMUM RATINGS

SYMBOL	PARAMETER	RATING	UNIT
V <sub>S</sub>	Supply voltage • A741C • A741	• 18	V
		• 22	V
P <sub>D</sub>	Internal power dissipation D package N package F package	780	mW
		1170	mW
		800	mW
V <sub>IN</sub>	Differential input voltage	• 30	V
V <sub>IN</sub>	Input voltage <sup>1</sup>	• 15	V
I <sub>SC</sub>	Output short-circuit duration	Continuous	
T <sub>A</sub>	Operating temperature range • A741C SA741C • A741	0 to +70	°C
		-40 to +85	°C
		-55 to +125	°C
T <sub>STG</sub>	Storage temperature range	-65 to +150	°C
T <sub>SOLD</sub>	Lead soldering temperature (10sec max)	300	°C

#### NOTES:

1. For supply voltages less than • 15V, the absolute maximum input voltage is equal to the supply voltage.

**OA741**

General purpose operational amplifier

• A741/• A741C/SA741C

DC ELECTRICAL CHARACTERISTICS

$T_A = 25^{\circ}\text{C}$ ,  $V_S = 15\text{V}$ , unless otherwise specified.

SYMBOL	PARAMETER	TEST CONDITIONS	• A741			• A741C			UNIT
			Min	Typ	Max	Min	Typ	Max	
$V_{OS}$ • $V_{OS}/T$	Offset voltage	$R_S=10k^{\bullet}$ $R_S=10k^{\bullet}$ , over temp.		1.0	5.0		2.0	6.0	mV
				1.0	6.0			7.5	mV
				10			10		
$I_{OS}$ • $I_{OS}/T$	Offset current	Over temp. $T_A=+125^{\bullet}\text{C}$ $T_A=-55^{\bullet}\text{C}$		20	200		20	200	nA
				7.0	200			300	nA
				20	500				nA
				200			200		
$I_{BIAS}$ • $I_B/T$	Input bias current	Over temp. $T_A=+125^{\bullet}\text{C}$ $T_A=-55^{\bullet}\text{C}$		80	500		80	500	nA
				30	500			800	nA
				300	1500				nA
				1			1		
$V_{OUT}$	Output voltage swing	$R_L=10k^{\bullet}$ $R_L=2k^{\bullet}$ , over temp.	• 12	• 14		• 12	• 14		V
			• 10	• 13		• 10	• 13		V
$A_{VOL}$	Large-signal voltage gain	$R_L=2k^{\bullet}$ , $V_O=10\text{V}$ $R_L=2k^{\bullet}$ , $V_O=10\text{V}$ , over temp.	50	200		20	200		V/mV
			25			15			V/mV
	Offset voltage adjustment range			• 30			• 30		mV
PSRR	Supply voltage rejection ratio	$R_S=10k^{\bullet}$ $R_S=10k^{\bullet}$ , over temp.					10	150	• V/V
									• V/V
CMRR	Common-mode rejection ratio	Over temp.				70	90		dB
			70	90					dB
$I_{CC}$	Supply current	$T_A=+125^{\bullet}\text{C}$ $T_A=-55^{\bullet}\text{C}$		1.4	2.8		1.4	2.8	mA
				1.5	2.5				mA
				2.0	3.3				mA
$V_{IN}$	Input voltage range	(• A741, over temp.)	• 12	• 13		• 12	• 13		V
$R_{IN}$	Input resistance		0.3	2.0		0.3	2.0		M•
$P_D$	Power consumption	$T_A=+125^{\bullet}\text{C}$ $T_A=-55^{\bullet}\text{C}$		50	85		50	85	mW
				45	75				mW
				45	100				mW
$R_{OUT}$	Output resistance			75			75		•
$I_{SC}$	Output short-circuit current		10	25	60	10	25	60	mA

**OA741**

## General purpose operational amplifier

• A741/• A741C/SA741C

## DC ELECTRICAL CHARACTERISTICS

 $T_A = 25^\circ\text{C}$ ,  $V_S = \pm 15\text{V}$ , unless otherwise specified.

SYMBOL	PARAMETER	TEST CONDITIONS	SA741C			UNIT
			Min	Typ	Max	
$V_{OS}$	Offset voltage	$R_S = 10k\Omega$		2.0	6.0	mV
$\cdot V_{OS}/\cdot T$		$R_S = 10k\Omega$ , over temp.		10	7.5	mV/ $^\circ\text{C}$
$I_{OS}$	Offset current	Over temp.		20	200	nA
$\cdot I_{OS}/\cdot T$				200	500	pA/ $^\circ\text{C}$
$I_{BIAS}$	Input bias current	Over temp.		80	500	nA
$\cdot I_B/\cdot T$				1	1500	nA/ $^\circ\text{C}$
$V_{OUT}$	Output voltage swing	$R_L = 10k\Omega$	$\cdot 12$	$\cdot 14$		V
		$R_L = 2k\Omega$ , over temp.	$\cdot 10$	$\cdot 13$		V
$A_{VOL}$	Large-signal voltage gain	$R_L = 2k\Omega$ , $V_O = \pm 10\text{V}$	20	200		V/mV
		$R_L = 2k\Omega$ , $V_O = \pm 10\text{V}$ , over temp.	15			V/mV
	Offset voltage adjustment range			$\cdot 30$		mV
PSRR	Supply voltage rejection ratio	$R_S = 10k\Omega$		10	150	$\cdot \text{V/V}$
CMRR	Common mode rejection ratio		70	90		dB
$V_{IN}$	Input voltage range	Over temp.	$\cdot 12$	$\cdot 13$		V
$R_{IN}$	Input resistance		0.3	2.0		M $\Omega$
$P_d$	Power consumption			50	85	mW
$R_{OUT}$	Output resistance			75		$\Omega$
$I_{SC}$	Output short-circuit current			25		mA

## AC ELECTRICAL CHARACTERISTICS

 $T_A = 25^\circ\text{C}$ ,  $V_S = \pm 15\text{V}$ , unless otherwise specified.

SYMBOL	PARAMETER	TEST CONDITIONS	• A741, • A741C			UNIT
			Min	Typ	Max	
$R_{IN}$	Parallel input resistance	Open-loop, $f = 20\text{Hz}$	0.3			M $\Omega$
$C_{IN}$	Parallel input capacitance	Open-loop, $f = 20\text{Hz}$		1.4		pF
	Unity gain crossover frequency	Open-loop		1.0		MHz
$t_R$	Transient response unity gain	$V_{IN} = 20\text{mV}$ , $R_L = 2k\Omega$ , $C_L = 100\text{pF}$		0.3		$\cdot \text{s}$
				5.0		%
SR	Slew rate	$C = 100\text{pF}$ , $R_L = 2k\Omega$ , $V_{IN} = \pm 10\text{V}$		0.5		V/ $\mu\text{s}$

**OA741**

General purpose operational amplifier

• A741/• A741C/SA741C

EQUIVALENT SCHEMATIC

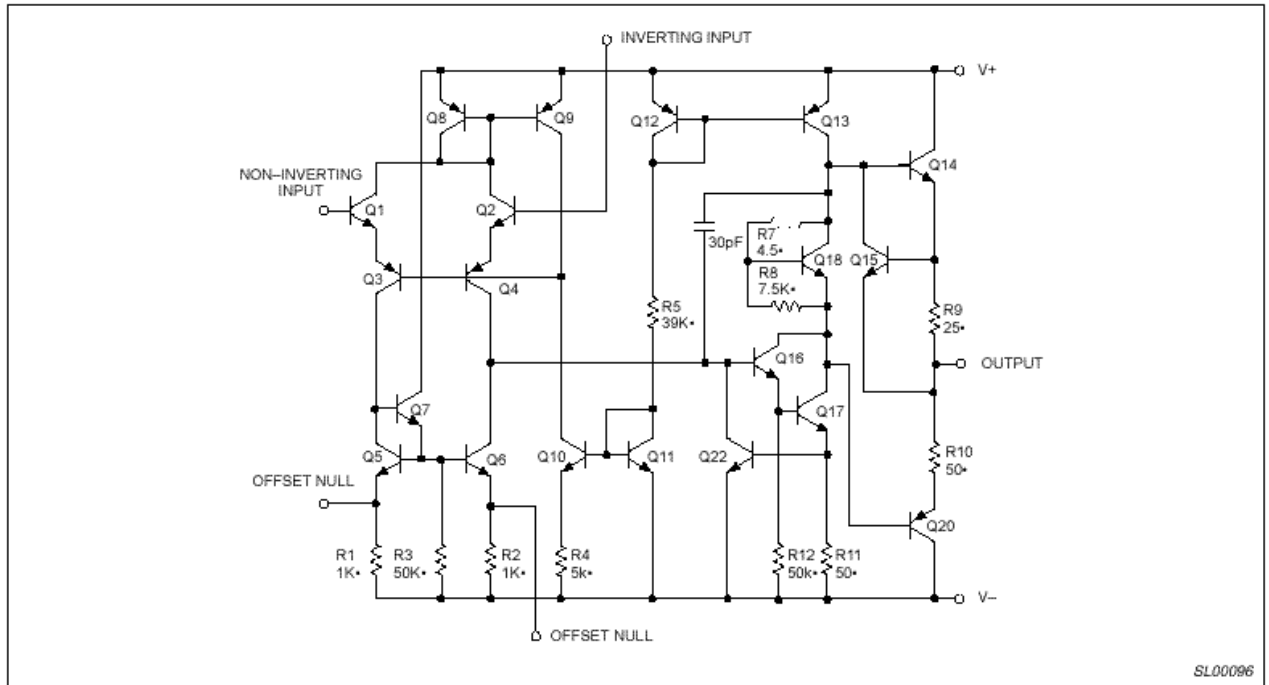


Figure 2. Equivalent Schematic

**OA741**

# General purpose operational amplifier

• A741 • A741C/SA741C

## TYPICAL PERFORMANCE CHARACTERISTICS

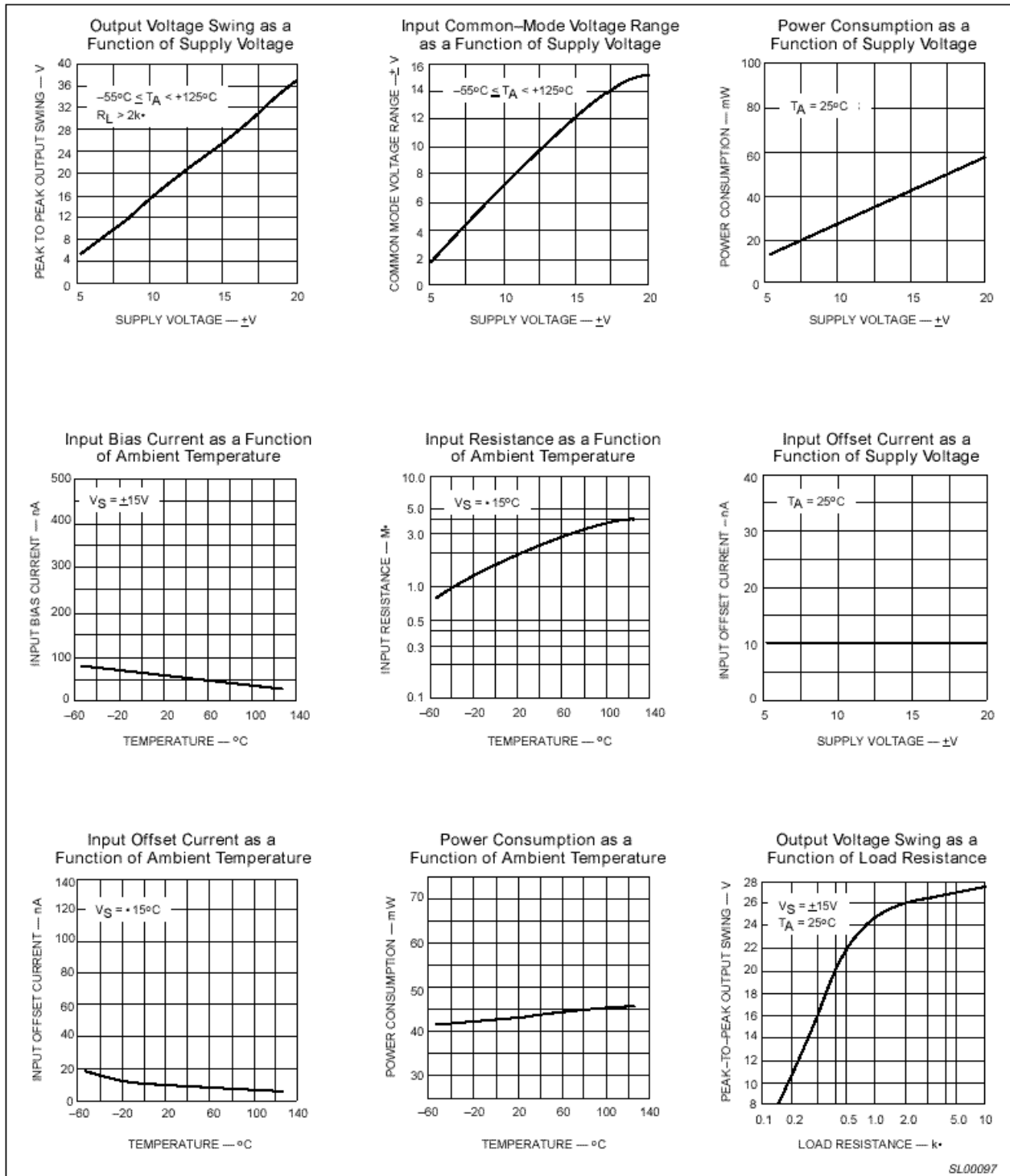


Figure 3. Typical Performance Characteristics

**OA741**

General purpose operational amplifier

• A741/• A741C/SA741C

TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

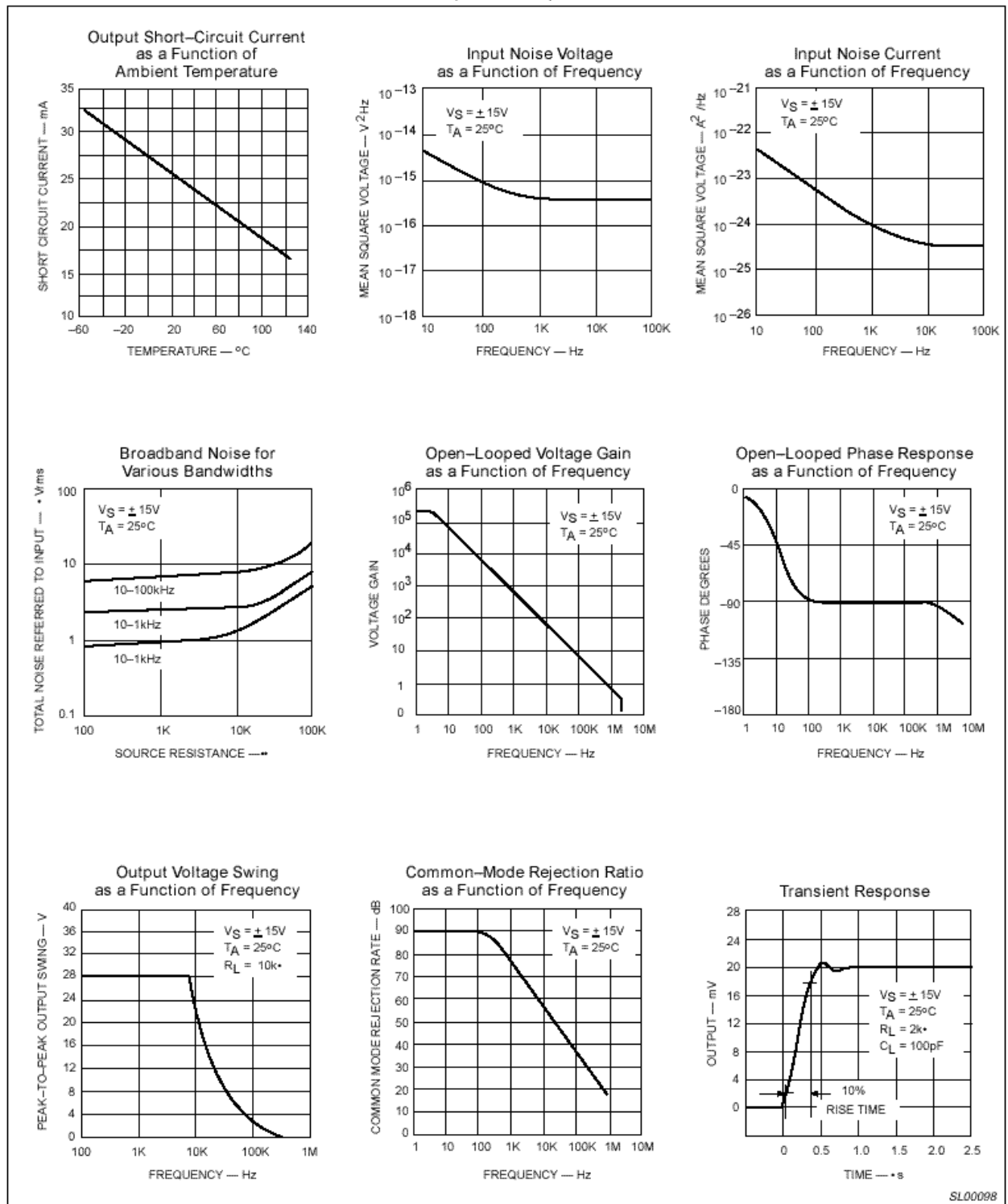
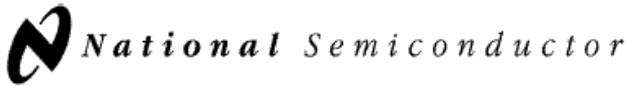


Figure 4. Typical Performance Characteristics (cont.)

**OA741**



August 1999

LM117/LM317A/LM317 3-Terminal Adjustable Regulator

## LM117/LM317A/LM317 3-Terminal Adjustable Regulator

### General Description

The LM117 series of adjustable 3-terminal positive voltage regulators is capable of supplying in excess of 1.5A over a 1.2V to 37V output range. They are exceptionally easy to use and require only two external resistors to set the output voltage. Further, both line and load regulation are better than standard fixed regulators. Also, the LM117 is packaged in standard transistor packages which are easily mounted and handled.

In addition to higher performance than fixed regulators, the LM117 series offers full overload protection available only in IC's. Included on the chip are current limit, thermal overload protection and safe area protection. All overload protection circuitry remains fully functional even if the adjustment terminal is disconnected.

Normally, no capacitors are needed unless the device is situated more than 6 inches from the input filter capacitors in which case an input bypass is needed. An optional output capacitor can be added to improve transient response. The adjustment terminal can be bypassed to achieve very high ripple rejection ratios which are difficult to achieve with standard 3-terminal regulators.

Besides replacing fixed regulators, the LM117 is useful in a wide variety of other applications. Since the regulator is "floating" and sees only the input-to-output differential volt-

age, supplies of several hundred volts can be regulated as long as the maximum input to output differential is not exceeded, i.e., avoid short-circuiting the output.

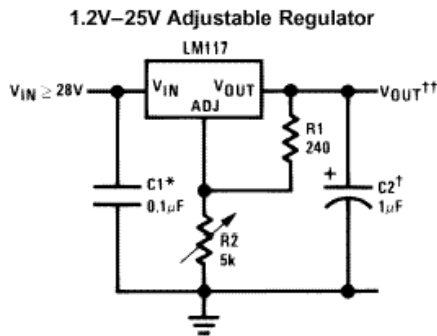
Also, it makes an especially simple adjustable switching regulator, a programmable output regulator, or by connecting a fixed resistor between the adjustment pin and output, the LM117 can be used as a precision current regulator. Supplies with electronic shutdown can be achieved by clamping the adjustment terminal to ground which programs the output to 1.2V where most loads draw little current.

For applications requiring greater output current, see LM150 series (3A) and LM138 series (5A) data sheets. For the negative complement, see LM137 series data sheet.

### Features

- Guaranteed 1% output voltage tolerance (LM317A)
- Guaranteed max. 0.01%/V line regulation (LM317A)
- Guaranteed max. 0.3% load regulation (LM117)
- Guaranteed 1.5A output current
- Adjustable output down to 1.2V
- Current limit constant with temperature
- P<sup>+</sup> Product Enhancement tested
- 80 dB ripple rejection
- Output is short-circuit protected

### Typical Applications



DS009063-1

Full output current not available at high input-output voltages  
 \*Needed if device is more than 6 inches from filter capacitors.

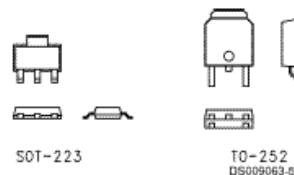
†Optional—improves transient response. Output capacitors in the range of 1 μF to 1000 μF of aluminum or tantalum electrolytic are commonly used to provide improved output impedance and rejection of transients.

$$V_{OUT} = 1.25V \left( 1 + \frac{R_2}{R_1} \right) + I_{ADJ}(R_2)$$

### LM117 Series Packages

Part Number Suffix	Package	Design Load Current
K	TO-3	1.5A
H	TO-39	0.5A
T	TO-220	1.5A
E	LCC	0.5A
S	TO-263	1.5A
EMP	SOT-223	1A
MDT	TO-252	0.5A

### SOT-223 vs D-Pak (TO-252) Packages



Scale 1:1



<p><b>Absolute Maximum Ratings</b> (Note 1)</p> <p>If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.</p> <p>Power Dissipation Internally Limited              Input-Output Voltage Differential +40V, -0.3V              Storage Temperature -65°C to +150°C              Lead Temperature                  Metal Package (Soldering, 10 seconds) 300°C                  Plastic Package (Soldering, 4 seconds) 260°C              ESD Tolerance (Note 5) 3 kV</p>		<p><b>Operating Temperature Range</b></p> <p>LM117 -55°C ≤ T<sub>J</sub> ≤ +150°C              LM317A -40°C ≤ T<sub>J</sub> ≤ +125°C              LM317 0°C ≤ T<sub>J</sub> ≤ +125°C</p>				
<p><b>Electrical Characteristics</b> (Note 3)</p> <p>Specifications with standard type face are for T<sub>J</sub> = 25°C, and those with <b>boldface type</b> apply over full Operating Temperature Range. Unless otherwise specified, V<sub>IN</sub> - V<sub>OUT</sub> = 5V, and I<sub>OUT</sub> = 10 mA.</p>		<p><b>Preconditioning</b></p> <p>Thermal Limit Burn-In All Devices 100%</p>				
Parameter	Conditions	LM117 (Note 2)			Units	
		Min	Typ	Max		
Reference Voltage					V	
	3V ≤ (V <sub>IN</sub> - V <sub>OUT</sub> ) ≤ 40V, 10 mA ≤ I <sub>OUT</sub> ≤ I <sub>MAX</sub> , P ≤ P <sub>MAX</sub>	<b>1.20</b>	<b>1.25</b>	<b>1.30</b>	V	
Line Regulation	3V ≤ (V <sub>IN</sub> - V <sub>OUT</sub> ) ≤ 40V (Note 4)		0.01	0.02	%/V	
			<b>0.02</b>	<b>0.05</b>	%/V	
Load Regulation	10 mA ≤ I <sub>OUT</sub> ≤ I <sub>MAX</sub> (Note 4)		0.1	0.3	%	
			<b>0.3</b>	<b>1</b>	%	
Thermal Regulation	20 ms Pulse		0.03	0.07	%/W	
Adjustment Pin Current			<b>50</b>	<b>100</b>	µA	
Adjustment Pin Current Change	10 mA ≤ I <sub>OUT</sub> ≤ I <sub>MAX</sub> 3V ≤ (V <sub>IN</sub> - V <sub>OUT</sub> ) ≤ 40V		<b>0.2</b>	<b>5</b>	µA	
Temperature Stability	T <sub>MIN</sub> ≤ T <sub>J</sub> ≤ T <sub>MAX</sub>		<b>1</b>		%	
Minimum Load Current	(V <sub>IN</sub> - V <sub>OUT</sub> ) = 40V		<b>3.5</b>	<b>5</b>	mA	
Current Limit	(V <sub>IN</sub> - V <sub>OUT</sub> ) ≤ 15V	K Package	<b>1.5</b>	<b>2.2</b>	<b>3.4</b>	A
		H Packages	<b>0.5</b>	<b>0.8</b>	<b>1.8</b>	A
	(V <sub>IN</sub> - V <sub>OUT</sub> ) = 40V	K Package	0.3	0.4		A
		H Package	0.15	0.2		A
RMS Output Noise, % of V <sub>OUT</sub>	10 Hz ≤ f ≤ 10 kHz		0.003		%	
Ripple Rejection Ratio	V <sub>OUT</sub> = 10V, f = 120 Hz, C <sub>ADJ</sub> = 0 µF		<b>65</b>		dB	
	V <sub>OUT</sub> = 10V, f = 120 Hz, C <sub>ADJ</sub> = 10 µF	<b>66</b>	<b>80</b>		dB	
Long-Term Stability	T <sub>J</sub> = 125°C, 1000 hrs		0.3	1	%	
Thermal Resistance, Junction-to-Case	K Package		2.3	3	°C/W	
	H Package		12	15	°C/W	
	E Package				°C/W	
Thermal Resistance, Junction-to-Ambient (No Heat Sink)	K Package		35		°C/W	
	H Package		140		°C/W	
	E Package				°C/W	

**LM117/317**

### Application Hints

In operation, the LM117 develops a nominal 1.25V reference voltage,  $V_{REF}$ , between the output and adjustment terminal. The reference voltage is impressed across program resistor R1 and, since the voltage is constant, a constant current  $I_1$  then flows through the output set resistor R2, giving an output voltage of

$$V_{OUT} = V_{REF} \left( 1 + \frac{R2}{R1} \right) + I_{ADJ}R2$$

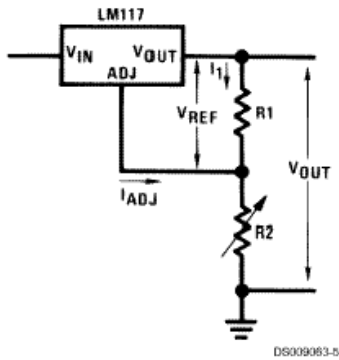


FIGURE 1.

Since the 100  $\mu$ A current from the adjustment terminal represents an error term, the LM117 was designed to minimize  $I_{ADJ}$  and make it very constant with line and load changes. To do this, all quiescent operating current is returned to the output establishing a minimum load current requirement. If there is insufficient load on the output, the output will rise.

#### External Capacitors

An input bypass capacitor is recommended. A 0.1  $\mu$ F disc or 1  $\mu$ F solid tantalum on the input is suitable input bypassing for almost all applications. The device is more sensitive to the absence of input bypassing when adjustment or output capacitors are used but the above values will eliminate the possibility of problems.

The adjustment terminal can be bypassed to ground on the LM117 to improve ripple rejection. This bypass capacitor prevents ripple from being amplified as the output voltage is increased. With a 10  $\mu$ F bypass capacitor 80 dB ripple rejection is obtainable at any output level. Increases over 10  $\mu$ F do not appreciably improve the ripple rejection at frequencies above 120 Hz. If the bypass capacitor is used, it is sometimes necessary to include protection diodes to prevent the capacitor from discharging through internal low current paths and damaging the device.

In general, the best type of capacitors to use is solid tantalum. Solid tantalum capacitors have low impedance even at high frequencies. Depending upon capacitor construction, it takes about 25  $\mu$ F in aluminum electrolytic to equal 1  $\mu$ F solid tantalum at high frequencies. Ceramic capacitors are also good at high frequencies; but some types have a large decrease in capacitance at frequencies around 0.5 MHz. For this reason, 0.01  $\mu$ F disc may seem to work better than a 0.1  $\mu$ F disc as a bypass.

Although the LM117 is stable with no output capacitors, like any feedback circuit, certain values of external capacitance can cause excessive ringing. This occurs with values between 500 pF and 5000 pF. A 1  $\mu$ F solid tantalum (or 25  $\mu$ F

aluminum electrolytic) on the output swamps this effect and insures stability. Any increase of the load capacitance larger than 10  $\mu$ F will merely improve the loop stability and output impedance.

#### Load Regulation

The LM117 is capable of providing extremely good load regulation but a few precautions are needed to obtain maximum performance. The current set resistor connected between the adjustment terminal and the output terminal (usually 240 $\Omega$ ) should be tied directly to the output (case) of the regulator rather than near the load. This eliminates line drops from appearing effectively in series with the reference and degrading regulation. For example, a 15V regulator with 0.05 $\Omega$  resistance between the regulator and load will have a load regulation due to line resistance of 0.05 $\Omega$  x  $I_L$ . If the set resistor is connected near the load the effective line resistance will be 0.05 $\Omega$  (1 + R2/R1) or in this case, 11.5 times worse.

Figure 2 shows the effect of resistance between the regulator and 240 $\Omega$  set resistor.

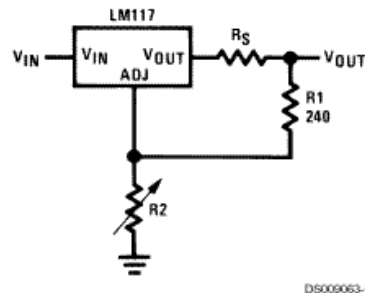


FIGURE 2. Regulator with Line Resistance in Output Lead

With the TO-3 package, it is easy to minimize the resistance from the case to the set resistor, by using two separate leads to the case. However, with the TO-39 package, care should be taken to minimize the wire length of the output lead. The ground of R2 can be returned near the ground of the load to provide remote ground sensing and improve load regulation.

#### Protection Diodes

When external capacitors are used with any IC regulator it is sometimes necessary to add protection diodes to prevent the capacitors from discharging through low current points into the regulator. Most 10  $\mu$ F capacitors have low enough internal series resistance to deliver 20A spikes when shorted. Although the surge is short, there is enough energy to damage parts of the IC.

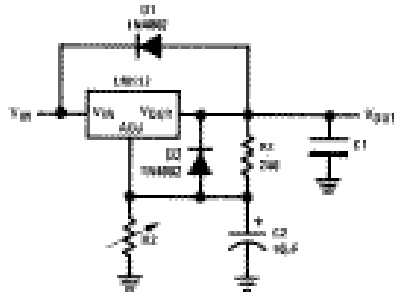
When an output capacitor is connected to a regulator and the input is shorted, the output capacitor will discharge into the output of the regulator. The discharge current depends on the value of the capacitor, the output voltage of the regulator, and the rate of decrease of  $V_{IN}$ . In the LM117, this discharge path is through a large junction that is able to sustain 15A surge with no problem. This is not true of other types of positive regulators. For output capacitors of 25  $\mu$ F or less, there is no need to use diodes.

The bypass capacitor on the adjustment terminal can discharge through a low current junction. Discharge occurs when either the input or output is shorted. Internal to the LM117 is a 50 $\Omega$  resistor which limits the peak discharge current. No protection is needed for output voltages of 25V or

**LM117/317**

**Application Hints** (Continued)

less and 10 µF capacitance. Figure 3 shows an LM117 with protection diodes included for use with outputs greater than 25V and high values of output capacitance.



$$V_{OUT} = 1.25V \left( 1 + \frac{R_2}{R_1} \right) + I_{AD}R_2$$

D1 protects against Q1  
D2 protects against Q2

**FIGURE 3. Regulator with Protection Diodes**

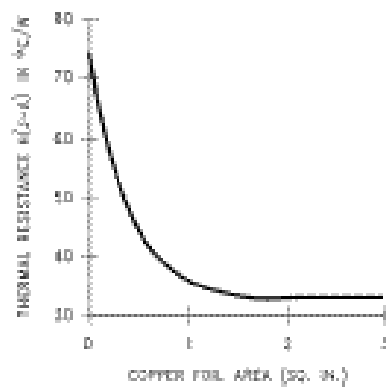
When a value for  $\theta_{(J-A)}$  is found using the equation shown, a heatsink must be selected that has a value that is less than or equal to this number.

$\theta_{(J-A)}$  is specified numerically by the heatsink manufacturer in the catalog, or shown in a curve that plots temperature rise vs power dissipation for the heatsink.

**HEATSINKING TO-263, SOT-223 AND TO-252 PACKAGE PARTS**

The TO-263 ("8"), SOT-223 ("MP") and TO-252 ("DT") packages use a copper plane on the PCB and the PCB itself as a heatsink. To optimize the heat sinking ability of the plane and PCB, solder the tab of the packages to the plane.

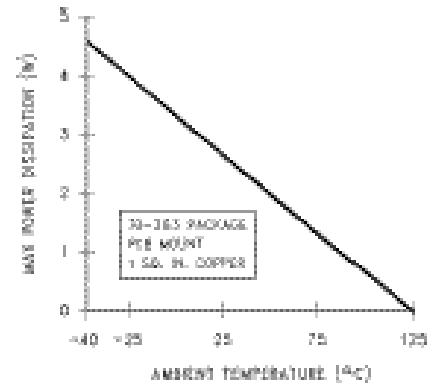
Figure 4 shows for the TO-263 the measured values of  $\theta_{(J-A)}$  for different copper area sizes using a typical PCB with 1 ounce copper and no solder mask over the copper area used for heatsinking.



**FIGURE 4.  $\theta_{(J-A)}$  vs Copper (1 ounce) Area for the TO-263 Package**

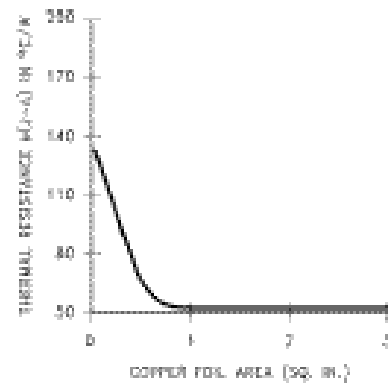
As shown in the figure, increasing the copper area beyond 1 square inch produces very little improvement. It should also be observed that the minimum value of  $\theta_{(J-A)}$  for the TO-263 package mounted to a PCB is 32°C/W.

As a design aid, Figure 5 shows the maximum allowable power dissipation compared to ambient temperature for the TO-263 device (assuming  $\theta_{(J-A)}$  is 35°C/W and the maximum junction temperature is 125°C).

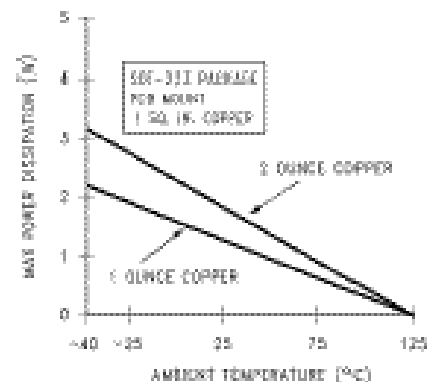


**FIGURE 5. Maximum Power Dissipation vs  $T_{Amb}$  for the TO-263 Package**

Figure 6 and Figure 7 show the information for the SOT-223 package. Figure 7 assumes a  $\theta_{(J-A)}$  of 74°C/W for 1 ounce copper and 51°C/W for 2 ounce copper and a maximum junction temperature of 125°C.

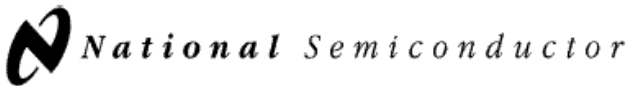


**FIGURE 6.  $\theta_{(J-A)}$  vs Copper (2 ounce) Area for the SOT-223 Package**



**FIGURE 7. Maximum Power Dissipation vs  $T_{Amb}$  for the SOT-223 Package**

**LM117/317**



May 1999

LM137/LM337 3-Terminal Adjustable Negative Regulators

# LM137/LM337 3-Terminal Adjustable Negative Regulators

## General Description

The LM137/LM337 are adjustable 3-terminal negative voltage regulators capable of supplying in excess of -1.5A over an output voltage range of -1.2V to -37V. These regulators are exceptionally easy to apply, requiring only 2 external resistors to set the output voltage and 1 output capacitor for frequency compensation. The circuit design has been optimized for excellent regulation and low thermal transients. Further, the LM137 series features internal current limiting, thermal shutdown and safe-area compensation, making them virtually blowout-proof against overloads.

The LM137/LM337 serve a wide variety of applications including local on-card regulation, programmable-output voltage regulation or precision current regulation. The LM137/LM337 are ideal complements to the LM117/LM317 adjustable positive regulators.

## Features

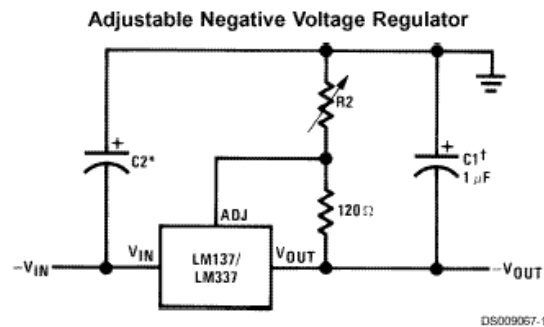
- Output voltage adjustable from -1.2V to -37V
- 1.5A output current guaranteed, -55°C to +150°C
- Line regulation typically 0.01%/V
- Load regulation typically 0.3%
- Excellent thermal regulation, 0.002%/W

- 77 dB ripple rejection
- Excellent rejection of thermal transients
- 50 ppm/°C temperature coefficient
- Temperature-independent current limit
- Internal thermal overload protection
- P\* Product Enhancement tested
- Standard 3-lead transistor package
- Output is short circuit protected

## LM137 Series Packages and Power Capability

Device	Package	Rated Power Dissipation	Design Load Current
LM137/337	TO-3 (K)	20W	1.5A
	TO-39 (H)	2W	0.5A
LM337	TO-220 (T)	15W	1.5A
LM337	SOT-223 (MP)	2W	1A

## Typical Applications



Full output current not available at high input-output voltages

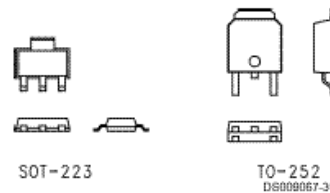
$$-V_{OUT} = -1.25V \left( 1 + \frac{R2}{120} \right) + (-I_{ADJ} \times R2)$$

†C1 = 1 μF solid tantalum or 10 μF aluminum electrolytic required for stability

\*C2 = 1 μF solid tantalum is required only if regulator is more than 4" from power-supply filter capacitor

Output capacitors in the range of 1 μF to 1000 μF of aluminum or tantalum electrolytic are commonly used to provide improved output impedance and rejection of transients

## Comparison between SOT-223 and D-Pak (TO-252) Packages



Scale 1:1

**LM137/337**

<b>Absolute Maximum Ratings</b> (Notes 1, 4)		LM337	0°C to +125°C	
If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.		Storage Temperature	-65°C to +150°C	
Power Dissipation	Internally Limited	Lead Temperature (Soldering, 10 sec.)	300°C	
Input-Output Voltage Differential	40V	Plastic Package (Soldering, 4 sec.)	260°C	
Operating Junction Temperature Range LM137	-55°C to +150°C	ESD Rating	2k Volts	

<b>Electrical Characteristics</b>	
(Note 1)	

Parameter	Conditions	LM137			LM337			Units
		Min	Typ	Max	Min	Typ	Max	
Line Regulation	$T_j = 25^\circ\text{C}$ , $3\text{V} \leq  V_{\text{IN}} - V_{\text{OUT}}  \leq 40\text{V}$ (Note 2) $I_L = 10\text{ mA}$		0.01	0.02		0.01	0.04	%/V
Load Regulation	$T_j = 25^\circ\text{C}$ , $10\text{ mA} \leq I_{\text{OUT}} \leq I_{\text{MAX}}$		0.3	0.5		0.3	1.0	%
Thermal Regulation	$T_j = 25^\circ\text{C}$ , 10 ms Pulse		0.002	0.02		0.003	0.04	%/W
Adjustment Pin Current			65	100		65	100	$\mu\text{A}$
Adjustment Pin Current Charge	$10\text{ mA} \leq I_L \leq I_{\text{MAX}}$ $3.0\text{V} \leq  V_{\text{IN}} - V_{\text{OUT}}  \leq 40\text{V}$ , $T_A = 25^\circ\text{C}$		2	5		2	5	$\mu\text{A}$
Reference Voltage	$T_j = 25^\circ\text{C}$ (Note 3) $3\text{V} \leq  V_{\text{IN}} - V_{\text{OUT}}  \leq 40\text{V}$ , (Note 3) $10\text{ mA} \leq I_{\text{OUT}} \leq I_{\text{MAX}}$ , $P \leq P_{\text{MAX}}$	-1.225	-1.250	-1.275	-1.213	-1.250	-1.287	V
		-1.200	-1.250	-1.300	-1.200	-1.250	-1.300	V
Line Regulation	$3\text{V} \leq  V_{\text{IN}} - V_{\text{OUT}}  \leq 40\text{V}$ , (Note 2)		0.02	0.05		0.02	0.07	%/V
Load Regulation	$10\text{ mA} \leq I_{\text{OUT}} \leq I_{\text{MAX}}$ , (Note 2)		0.3	1		0.3	1.5	%
Temperature Stability	$T_{\text{MIN}} \leq T_j \leq T_{\text{MAX}}$		0.6			0.6		%
Minimum Load Current	$ V_{\text{IN}} - V_{\text{OUT}}  \leq 40\text{V}$ $ V_{\text{IN}} - V_{\text{OUT}}  \leq 10\text{V}$		2.5	5		2.5	10	mA
			1.2	3		1.5	6	mA
Current Limit	$ V_{\text{IN}} - V_{\text{OUT}}  \leq 15\text{V}$ K, MP and T Package	1.5	2.2	3.5	1.5	2.2	3.7	A
		0.5	0.8	1.8	0.5	0.8	1.9	A
	$ V_{\text{IN}} - V_{\text{OUT}}  = 40\text{V}$ , $T_j = 25^\circ\text{C}$	0.24	0.4		0.15	0.4		A
		0.15	0.17		0.10	0.17		A
RMS Output Noise, % of $V_{\text{OUT}}$	$T_j = 25^\circ\text{C}$ , $10\text{ Hz} \leq f \leq 10\text{ kHz}$		0.003			0.003		%
Ripple Rejection Ratio	$V_{\text{OUT}} = -10\text{V}$ , $f = 120\text{ Hz}$ $C_{\text{ADJ}} = 10\text{ }\mu\text{F}$		60			60		dB
		66	77		66	77		dB
Long-Term Stability	$T_j = 125^\circ\text{C}$ , 1000 Hours		0.3	1		0.3	1	%
Thermal Resistance, Junction to Case	H Package		12	15		12	15	$^\circ\text{C/W}$
	K Package		2.3	3		2.3	3	$^\circ\text{C/W}$
	T Package					4		$^\circ\text{C/W}$
Thermal Resistance, Junction to Ambient (No Heat Sink)	H Package		140			140		$^\circ\text{C/W}$
	K Package		35			35		$^\circ\text{C/W}$
	T Package					50		$^\circ\text{C/W}$
	MP Package					170		$^\circ\text{C/W}$

**Note 1:** Unless otherwise specified, these specifications apply  $-55^\circ\text{C} \leq T_j \leq +150^\circ\text{C}$  for the LM137,  $0^\circ\text{C} \leq T_j \leq +125^\circ\text{C}$  for the LM337;  $V_{\text{IN}} - V_{\text{OUT}} = 5\text{V}$ ; and  $I_{\text{OUT}} = 0.1\text{A}$  for the TO-39 package and  $I_{\text{OUT}} = 0.5\text{A}$  for the TO-3, SOT-223 and TO-220 packages. Although power dissipation is internally limited, these specifications are applicable for power dissipations of 2W for the TO-39 and SOT-223 (see Application Hints), and 20W for the TO-3, and TO-220.  $I_{\text{MAX}}$  is 1.5A for the TO-3, SOT-223 and TO-220 packages, and 0.2A for the TO-39 package.

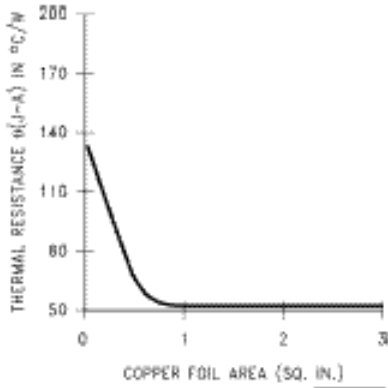
**Note 2:** Regulation is measured at constant junction temperature, using pulse testing with a low duty cycle. Changes in output voltage due to heating effects are covered under the specification for thermal regulation. Load regulation is measured on the output pin at a point  $\frac{1}{8}$ " below the base of the TO-3 and TO-39 packages.

**Note 3:** Selected devices with tightened tolerance reference voltage available.

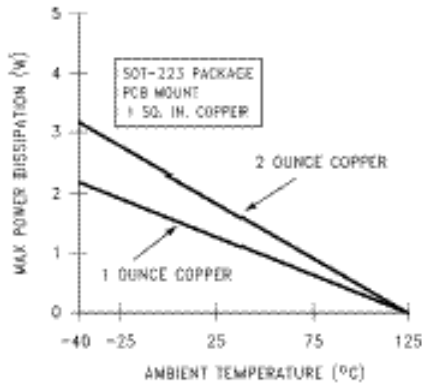
**Note 4:** Refer to RETS137H drawing for LM137H or RETS137K drawing for LM137K military specifications.

**LM137/337**

**Application Hints (Continued)**



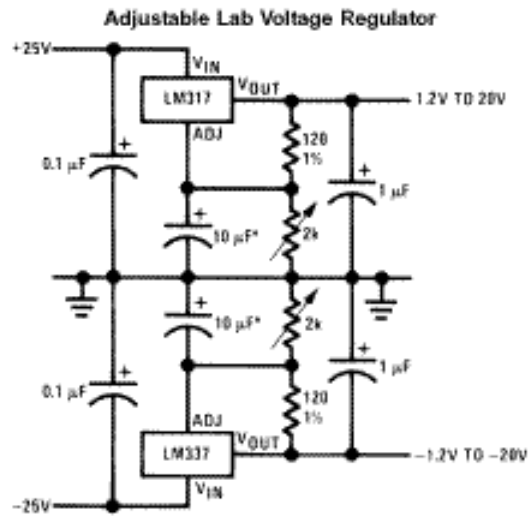
**FIGURE 3.  $\theta_{(J-A)}$  vs Copper (2 ounce) Area for the SOT-223 Package**



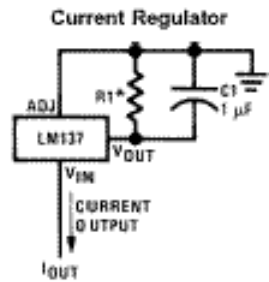
**FIGURE 4. Maximum Power Dissipation vs  $T_{AMB}$  for the SOT-223 Package**

Please see AN1028 for power enhancement techniques to be used with the SOT-223 package.

**Typical Applications**



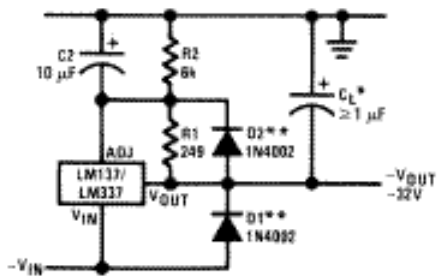
Full output current not available at high input-output voltages  
 \*The 10  $\mu\text{F}$  capacitors are optional to improve ripple rejection



$$I_{OUT} = \frac{1.250V}{R1}$$

\* $0.801 \leq R1 \leq 120\Omega$

**Negative Regulator with Protection Diodes**



\*\*When  $C_L$  is larger than 20  $\mu\text{F}$ , D1 protects the LM137 in case the input supply is shorted  
 \*\*When  $C_2$  is larger than 10  $\mu\text{F}$  and  $-V_{OUT}$  is larger than  $-25V$ , D2 protects the LM137 in case the output is shorted


# MKP3V120, MKP3V240

Preferred Device

## Sidac High Voltage

### Bidirectional Triggers

Bidirectional devices designed for direct interface with the ac power line. Upon reaching the breakover voltage in each direction, the device switches from a blocking state to a low voltage on-state. Conduction will continue like a Triac until the main terminal current drops below the holding current. The plastic axial lead package provides high pulse current capability at low cost. Glass passivation insures reliable operation. Applications are:

- High Pressure Sodium Vapor Lighting
- Strobes and Flashers
- Ignitors
- High Voltage Regulators
- Pulse Generators
- Used to Trigger Gates of SCR's and Triacs
-  Indicates UL Registered — File #E116110
- Device Marking: Logo, Device Type, e.g., MKP3V120, Date Code


#### MAXIMUM RATINGS (T<sub>J</sub> = 25°C unless otherwise noted)

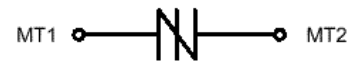
Rating	Symbol	Value	Unit
Peak Repetitive Off-State Voltage (Sine Wave, 50 to 60 Hz, T <sub>J</sub> = -40 to 125°C) MKP3V120 MKP3V240	V <sub>DRM</sub> , V <sub>RRM</sub>	±90 ±180	Volts
On-State RMS Current (T <sub>L</sub> = 80°C, Lead Length = 3/8", All Conduction Angles)	I <sub>T(RMS)</sub>	±1.0	Amp
Peak Non-Repetitive Surge Current (60 Hz One Cycle Sine Wave, Peak Value, T <sub>J</sub> = 125°C)	I <sub>TSM</sub>	±20	Amps
Operating Junction Temperature Range	T <sub>J</sub>	-40 to +125	°C
Storage Temperature Range	T <sub>stg</sub>	-40 to +150	°C



**ON Semiconductor**

<http://onsemi.com>

**SIDACS (  )  
1 AMPERE RMS  
120 and 240 VOLTS**



**SURMETIC 50  
PLASTIC AXIAL  
(No Polarity)  
CASE 267  
STYLE 2**

#### ORDERING INFORMATION

Device	Package	Shipping
MKP3V120	SURMETIC 50	Bulk 500/Bag
MKP3V120RL	SURMETIC 50	Tape and Reel 1.5K/Reel
MKP3V240	SURMETIC 50	Bulk 500/Bag
MKP3V240RL	SURMETIC 50	Tape and Reel 1.5K/Reel

Preferred devices are recommended choices for future use and best overall value.

### MKP3V129/240

### MKP3V120, MKP3V240

#### THERMAL CHARACTERISTICS

Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Lead (Lead Length = 3/8")	$R_{\theta JL}$	15	°C/W
Lead Solder Temperature (Lead Length $\geq$ 1/16" from Case, 10 s Max)	$T_L$	260	°C

#### ELECTRICAL CHARACTERISTICS ( $T_C = 25^\circ\text{C}$ unless otherwise noted; Electricals apply in both directions)

Characteristic	Symbol	Min	Typ	Max	Unit
----------------	--------	-----	-----	-----	------

#### OFF CHARACTERISTICS

Repetitive Peak Off-State Current (50 to 60 Hz Sine Wave) $V_{DRM} = 90\text{ V}$ $V_{DRM} = 180\text{ V}$	$I_{DRM}$	—	—	10	$\mu\text{A}$
MKP3V120 MKP3V240					

#### ON CHARACTERISTICS

Breakover Voltage, $I_{BO} = 200\ \mu\text{A}$	$V_{BO}$	110 220	— —	130 250	Volts
MKP3V120 MKP3V240					
Breakover Current	$I_{BO}$	—	—	200	$\mu\text{A}$
Peak On-State Voltage ( $I_{TM} = 1\text{ A Peak}$ , Pulse Width $\leq 300\ \mu\text{s}$ , Duty Cycle $\leq 2\%$ )	$V_{TM}$	—	1.1	1.5	Volts
Dynamic Holding Current (Sine Wave, 60 Hz, $R_L = 100\ \Omega$ )	$I_H$	—	—	100	mA
Switching Resistance (Sine Wave, 50 to 60 Hz)	$R_S$	0.1	—	—	k $\Omega$

#### DYNAMIC CHARACTERISTICS

Critical Rate-of-Rise of On-State Current, Critical Damped Waveform Circuit ( $I_{PK} = 130\text{ Amps}$ , Pulse Width = 10 $\mu\text{sec}$ )	$di/dt$	—	120	—	A/ $\mu\text{s}$
---	---------	---	-----	---	------------------

**MKP3V129/240**



**MOTOROLA**  
SEMICONDUCTOR TECHNICAL DATA

Order this document  
by MBS4991/D

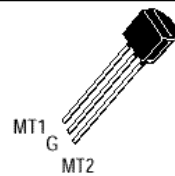
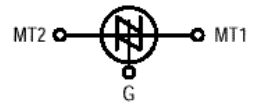
**Silicon Bidirectional Switches**  
Diode Thyristors

... designed for full-wave triggering in Triac phase control circuits, half-wave SCR triggering application and as voltage level detectors. Supplied in an inexpensive plastic TO-226AA package for high-volume requirements, this low-cost plastic package is readily adaptable for use in automatic insertion equipment.

- Low Switching Voltage — 8 Volts Typical
- Uniform Characteristics in Each Direction
- Low On-State Voltage — 1.7 Volts Maximum
- Low Off-State Current — 0.1  $\mu$ A Maximum
- Low Temperature Coefficient — 0.02 %/°C Typical

**MBS4991**  
**MBS4992**  
**MBS4993**

**SBS**  
**(PLASTIC)**



**CASE 29-04**  
**(TO-226AA)**  
**STYLE 12**

**MAXIMUM RATINGS** ( $T_J = 25^\circ\text{C}$  unless otherwise noted.)

Rating	Symbol	Value	Unit
Power Dissipation	$P_D$	500	mW
DC Forward Current	$I_F$	200	mA
DC Gate Current (Off-State Only)	$I_{G(\text{off})}$	5	mA
Repetitive Peak Forward Current (1% Duty Cycle, 10 $\mu$ s Pulse Width, $T_A = 100^\circ\text{C}$ )	$I_{FM(\text{rep})}$	2	Amps
Non-repetitive Forward Current (10 $\mu$ s Pulse Width, $T_A = 25^\circ\text{C}$ )	$I_{FM(\text{nonrep})}$	6	Amps
Operating Junction Temperature Range	$T_J$	-55 to +125	$^\circ\text{C}$
Storage Temperature Range	$T_{\text{stg}}$	-65 to +150	$^\circ\text{C}$

**MBS4991/2/3**

**MBS4991 MBS4992 MBS4993****ELECTRICAL CHARACTERISTICS** ( $T_A = 25^\circ\text{C}$  unless otherwise noted.)

Characteristic		Symbol	Min	Typ	Max	Unit
Switching Voltage	MBS4991 MBS4992, MBS4993	$V_S$	6 7.5	8 8	10 9	Vdc
Switching Current	MBS4991 MBS4992 MBS4993	$I_S$	— —	175 90 175	500 120 250	$\mu\text{Adc}$
Switching Voltage Differential (See Figure 10)	MBS4991 MBS4992, MBS4993	$ V_{S1}-V_{S2} $	— —	0.3 0.1	0.5 0.2	Vdc
Gate Trigger Current ( $V_F = 5 \text{ Vdc}$ , $R_L = 1 \text{ k ohm}$ )	MBS4992 MBS4993	$I_{GF}$	— —	— —	100 500	$\mu\text{Adc}$
Holding Current	MBS4991 MBS4992 MBS4993	$I_H$	— — —	0.7 0.2 0.3	1.5 0.5 0.75	mAdc
Off-State Blocking Current ( $V_F = 5 \text{ Vdc}$ , $T_A = 25^\circ\text{C}$ ) ( $V_F = 5 \text{ Vdc}$ , $T_A = 85^\circ\text{C}$ ) ( $V_F = 5 \text{ Vdc}$ , $T_A = 25^\circ\text{C}$ ) ( $V_F = 5 \text{ Vdc}$ , $T_A = 100^\circ\text{C}$ )	MBS4991 MBS4991 MBS4992, MBS4993 MBS4992, MBS4993	$I_B$	— — — —	0.08 2 0.08 6	1 10 0.1 10	$\mu\text{Adc}$
Forward On-State Voltage ( $I_F = 175 \text{ mAdc}$ ) ( $I_F = 200 \text{ mAdc}$ )	MBS4991 MBS4992, MBS4993	$V_F$	— —	1.4 1.5	1.7 1.7	Vdc
Peak Output Voltage ( $C_C = 0.1 \mu\text{F}$ , $R_L = 20 \text{ ohms}$ , (Figure 7))		$V_O$	3.5	4.8	—	Vdc
Turn-On Time (Figure 8)		$t_{on}$	—	1	—	$\mu\text{s}$
Turn-Off Time (Figure 9)		$t_{off}$	—	30	—	$\mu\text{s}$
Temperature Coefficient of Switching Voltage ( $-50$ to $+125^\circ\text{C}$ )		$T_C$	—	+0.02	—	$\%/^\circ\text{C}$
Switching Current Differential (See Figure 10)		$ I_{S1}-I_{S2} $	—	—	100	$\mu\text{A}$

**MBS4991/2/3**

# 2N5060 Series

Preferred Device

## Sensitive Gate Silicon Controlled Rectifiers

### Reverse Blocking Thyristors

Annular PNP devices designed for high volume consumer applications such as relay and lamp drivers, small motor controls, gate drivers for larger thyristors, and sensing and detection circuits. Supplied in an inexpensive plastic TO-226AA (TO-92) package which is readily adaptable for use in automatic insertion equipment.

- Sensitive Gate Trigger Current — 200  $\mu$ A Maximum
- Low Reverse and Forward Blocking Current — 50  $\mu$ A Maximum,  $T_C = 110^\circ\text{C}$
- Low Holding Current — 5 mA Maximum
- Passivated Surface for Reliability and Uniformity
- Device Marking: Device Type, e.g., 2N5060, Date Code

**MAXIMUM RATINGS** ( $T_J = 25^\circ\text{C}$  unless otherwise noted)

Rating	Symbol	Value	Unit
Peak Repetitive Off-State Voltage <sup>(1)</sup> ( $T_J = -40$ to $110^\circ\text{C}$ , Sine Wave, 50 to 60 Hz, Gate Open)	$V_{DRM}$ , $V_{RRM}$	30 60 100 200	Volts
On-State Current RMS ( $180^\circ$ Conduction Angles; $T_C = 80^\circ\text{C}$ )	$I_T(\text{RMS})$	0.8	Amp
*Average On-State Current ( $180^\circ$ Conduction Angles) ( $T_C = 67^\circ\text{C}$ ) ( $T_C = 102^\circ\text{C}$ )	$I_T(\text{AV})$	0.51 0.255	Amp
*Peak Non-repetitive Surge Current, $T_A = 25^\circ\text{C}$ (1/2 cycle, Sine Wave, 60 Hz)	$I_{TSM}$	10	Amps
Circuit Fusing Considerations ( $t = 8.3$ ms)	$I^2t$	0.4	$\text{A}^2\text{s}$
*Forward Peak Gate Power (Pulse Width $\leq 1.0$ $\mu\text{sec}$ ; $T_A = 25^\circ\text{C}$ )	$P_{GM}$	0.1	Watt
*Forward Average Gate Power ( $T_A = 25^\circ\text{C}$ , $t = 8.3$ ms)	$P_{G(\text{AV})}$	0.01	Watt
*Forward Peak Gate Current (Pulse Width $\leq 1.0$ $\mu\text{sec}$ ; $T_A = 25^\circ\text{C}$ )	$I_{GM}$	1.0	Amp
*Reverse Peak Gate Voltage (Pulse Width $\leq 1.0$ $\mu\text{sec}$ ; $T_A = 25^\circ\text{C}$ )	$V_{RGM}$	5.0	Volts
*Operating Junction Temperature Range	$T_J$	-40 to +110	$^\circ\text{C}$
*Storage Temperature Range	$T_{\text{stg}}$	-40 to +150	$^\circ\text{C}$

\*Indicates JEDEC Registered Data.

(1)  $V_{DRM}$  and  $V_{RRM}$  for all types can be applied on a continuous basis. Ratings apply for zero or negative gate voltage; however, positive gate voltage shall not be applied concurrent with negative potential on the anode. Blocking voltages shall not be tested with a constant current source such that the voltage ratings of the devices are exceeded.

### 2N5060



**ON Semiconductor**

<http://onsemi.com>

**SCRs**  
**0.8 AMPERES RMS**  
**30 thru 200 VOLTS**



TO-92 (TO-226AA)  
CASE 029  
STYLE 10

PIN ASSIGNMENT	
1	Cathode
2	Gate
3	Anode

### ORDERING INFORMATION

See detailed ordering and shipping information in the package dimensions section on page 7 of this data sheet.

Preferred devices are recommended choices for future use and best overall value.

## 2N5060 Series

### THERMAL CHARACTERISTICS

Characteristic	Symbol	Max	Unit
*Thermal Resistance, Junction to Case <sup>(1)</sup>	$R_{\theta JC}$	75	°C/W
Thermal Resistance, Junction to Ambient	$R_{\theta JA}$	200	°C/W
*Lead Solder Temperature (Lead Length $\geq$ 1/16" from case, 10 s Max)	—	+230*	°C

### ELECTRICAL CHARACTERISTICS ( $T_C = 25^\circ\text{C}$ unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
----------------	--------	-----	-----	-----	------

### OFF CHARACTERISTICS

*Peak Repetitive Forward or Reverse Blocking Current <sup>(2)</sup> ( $V_{AK} = \text{Rated } V_{DRM} \text{ or } V_{RRM}$ )	$I_{DRM}, I_{RRM}$	—	—	10	$\mu\text{A}$
$T_C = 25^\circ\text{C}$		—	—	50	$\mu\text{A}$
$T_C = 110^\circ\text{C}$		—	—	—	—

### ON CHARACTERISTICS

*Peak Forward On-State Voltage <sup>(3)</sup> ( $I_{TM} = 1.2 \text{ A peak @ } T_A = 25^\circ\text{C}$ )	$V_{TM}$	—	—	1.7	Volts
Gate Trigger Current (Continuous dc) <sup>(4)</sup> *( $V_{AK} = 7 \text{ Vdc}, R_L = 100 \text{ Ohms}$ )	$I_{GT}$	—	—	200	$\mu\text{A}$
$T_C = 25^\circ\text{C}$		—	—	350	
$T_C = -40^\circ\text{C}$		—	—	—	
Gate Trigger Voltage (Continuous dc) <sup>(4)</sup> *( $V_{AK} = 7 \text{ Vdc}, R_L = 100 \text{ Ohms}$ )	$V_{GT}$	—	—	0.8	Volts
$T_C = 25^\circ\text{C}$		—	—	1.2	
$T_C = -40^\circ\text{C}$		—	—	—	
*Gate Non-Trigger Voltage ( $V_{AK} = \text{Rated } V_{DRM}, R_L = 100 \text{ Ohms}$ )	$V_{GD}$	0.1	—	—	Volts
$T_C = 110^\circ\text{C}$		—	—	—	
Holding Current <sup>(4)</sup> *( $V_{AK} = 7 \text{ Vdc}, \text{ initiating current} = 20 \text{ mA}$ )	$I_H$	—	—	5.0	mA
$T_C = 25^\circ\text{C}$		—	—	10	
$T_C = -40^\circ\text{C}$		—	—	—	
Turn-On Time Delay Time Rise Time ( $I_{GT} = 1 \text{ mA}, V_D = \text{Rated } V_{DRM},$ Forward Current = 1 A, $di/dt = 6 \text{ A}/\mu\text{s}$ )	$t_d$ $t_r$	—	3.0 0.2	—	$\mu\text{s}$
Turn-Off Time (Forward Current = 1 A pulse, Pulse Width = 50 $\mu\text{s}$ , 0.1% Duty Cycle, $di/dt = 6 \text{ A}/\mu\text{s}$ , $dv/dt = 20 \text{ V}/\mu\text{s}, I_{GT} = 1 \text{ mA}$ )	$t_q$	—	10	—	$\mu\text{s}$
		—	30	—	
		—	—	—	

### DYNAMIC CHARACTERISTICS

Critical Rate of Rise of Off-State Voltage (Rated $V_{DRM}$ , Exponential)	$dv/dt$	—	30	—	$\text{V}/\mu\text{s}$
---	---------	---	----	---	------------------------

\*Indicates JEDEC Registered Data.

- (1) This measurement is made with the case mounted "flat side down" on a heat sink and held in position by means of a metal clamp over the curved surface.
- (2)  $R_{GK} = 1000 \Omega$  is included in measurement.
- (3) Forward current applied for 1 ms maximum duration, duty cycle  $\leq$  1%.
- (4)  $R_{GK}$  current is not included in measurement.



# 2N6027, 2N6028

Preferred Device

## Programmable Unijunction Transistor

### Programmable Unijunction Transistor Triggers

Designed to enable the engineer to “program” unijunction characteristics such as  $R_{BB}$ ,  $\eta$ ,  $I_V$ , and  $I_P$  by merely selecting two resistor values. Application includes thyristor–trigger, oscillator, pulse and timing circuits. These devices may also be used in special thyristor applications due to the availability of an anode gate. Supplied in an inexpensive TO–92 plastic package for high–volume requirements, this package is readily adaptable for use in automatic insertion equipment.

- Programmable —  $R_{BB}$ ,  $\eta$ ,  $I_V$  and  $I_P$
- Low On–State Voltage — 1.5 Volts Maximum @  $I_F = 50$  mA
- Low Gate to Anode Leakage Current — 10 nA Maximum
- High Peak Output Voltage — 11 Volts Typical
- Low Offset Voltage — 0.35 Volt Typical ( $R_G = 10$  k ohms)
- Device Marking: Logo, Device Type, e.g., 2N6027, Date Code

#### MAXIMUM RATINGS ( $T_J = 25^\circ\text{C}$ unless otherwise noted)

Rating	Symbol	Value	Unit
*Power Dissipation Derate Above $25^\circ\text{C}$	$P_F$ $1/\theta_{JA}$	300 4.0	mW mW/ $^\circ\text{C}$
*DC Forward Anode Current Derate Above $25^\circ\text{C}$	$I_T$	150 2.67	mA mA/ $^\circ\text{C}$
*DC Gate Current	$I_G$	$\pm 50$	mA
Repetitive Peak Forward Current 100 $\mu\text{s}$ Pulse Width, 1% Duty Cycle *20 $\mu\text{s}$ Pulse Width, 1% Duty Cycle	$I_{TRM}$	1.0 2.0	Amps
Non–Repetitive Peak Forward Current 10 $\mu\text{s}$ Pulse Width	$I_{TSM}$	5.0	Amps
*Gate to Cathode Forward Voltage	$V_{GKF}$	40	Volts
*Gate to Cathode Reverse Voltage	$V_{GKR}$	– 5.0	Volts
*Gate to Anode Reverse Voltage	$V_{GAR}$	40	Volts
*Anode to Cathode Voltage <sup>(1)</sup>	$V_{AK}$	$\pm 40$	Volts
Operating Junction Temperature Range	$T_J$	–50 to +100	$^\circ\text{C}$
*Storage Temperature Range	$T_{stg}$	–55 to +150	$^\circ\text{C}$

\*Indicates JEDEC Registered Data

(1) Anode positive,  $R_{GA} = 1000$  ohms  
Anode negative,  $R_{GA} = \text{open}$

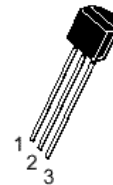
## 2N6027/28



ON Semiconductor

<http://onsemi.com>

PUTs  
40 VOLTS  
300 mW



TO–92 (TO–226AA)  
CASE 029  
STYLE 16

#### PIN ASSIGNMENT

1	Anode
2	Gate
3	Cathode

#### ORDERING INFORMATION

See detailed ordering and shipping information in the package dimensions section on page 7 of this data sheet.

**Preferred** devices are recommended choices for future use and best overall value.

**THERMAL CHARACTERISTICS**

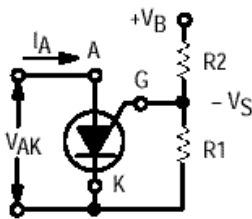
Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Case	$R_{\theta JC}$	75	$^{\circ}C/W$
Thermal Resistance, Junction to Ambient	$R_{\theta JA}$	200	$^{\circ}C/W$
Maximum Lead Temperature for Soldering Purposes ( $\leq 1/16''$ from case, 10 secs max)	$T_L$	260	$^{\circ}C$

**ELECTRICAL CHARACTERISTICS** ( $T_C = 25^{\circ}C$  unless otherwise noted.)

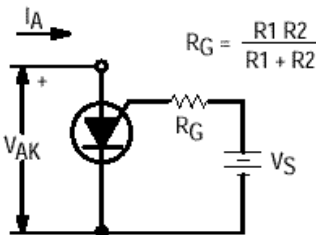
Characteristic	Fig. No.	Symbol	Min	Typ	Max	Unit
*Peak Current ( $V_S = 10$ Vdc, $R_G = 1$ M $\Omega$ ) 2N6027 2N6028 ( $V_S = 10$ Vdc, $R_G = 10$ k ohms) 2N6027 2N6028	2,9,11	$I_P$	—	1.25 0.08 4.0 0.70	2.0 0.15 5.0 1.0	$\mu A$
*Offset Voltage ( $V_S = 10$ Vdc, $R_G = 1$ M $\Omega$ ) 2N6027 2N6028 ( $V_S = 10$ Vdc, $R_G = 10$ k ohms) (Both Types)	1	$V_T$	0.2 0.2 0.2	0.70 0.50 0.35	1.6 0.6 0.6	Volts
*Valley Current ( $V_S = 10$ Vdc, $R_G = 1$ M $\Omega$ ) 2N6027 2N6028 ( $V_S = 10$ Vdc, $R_G = 10$ k ohms) 2N6027 2N6028 ( $V_S = 10$ Vdc, $R_G = 200$ ohms) 2N6027 2N6028	1,4,5	$I_V$	— — 70 25 1.5 1.0	18 18 150 150 — —	50 25 — — — —	$\mu A$     mA
*Gate to Anode Leakage Current ( $V_S = 40$ Vdc, $T_A = 25^{\circ}C$ , Cathode Open) ( $V_S = 40$ Vdc, $T_A = 75^{\circ}C$ , Cathode Open)	—	$I_{GAO}$	— —	1.0 3.0	10 —	nAdc
Gate to Cathode Leakage Current ( $V_S = 40$ Vdc, Anode to Cathode Shorted)	—	$I_{GKS}$	—	5.0	50	nAdc
*Forward Voltage ( $I_F = 50$ mA Peak)(1)	1,6	$V_F$	—	0.8	1.5	Volts
*Peak Output Voltage ( $V_G = 20$ Vdc, $C_C = 0.2$ $\mu F$ )	3,7	$V_O$	6.0	11	—	Volt
Pulse Voltage Rise Time ( $V_B = 20$ Vdc, $C_C = 0.2$ $\mu F$ )	3	$t_r$	—	40	80	ns

\*Indicates JEDEC Registered Data

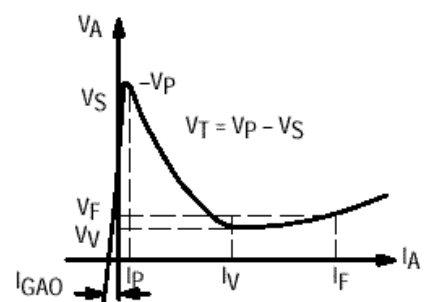
(1) Pulse Test: Pulse Width  $\leq 300$   $\mu$ sec, Duty Cycle  $\leq 2\%$ .



1A - Programmable Unijunction with "Program" Resistors R1 and R2



1B - Equivalent Test Circuit for Figure 1A used for electrical characteristics testing (also see Figure 2)



IC - Electrical Characteristics

**2N6027/28**



2N2646

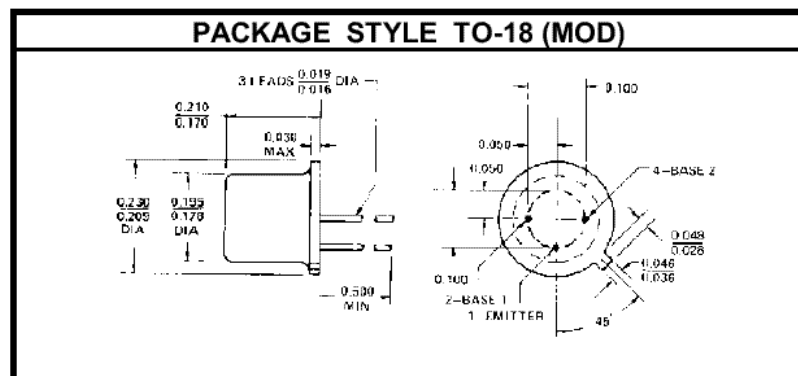
# SILICON PN UNIJUNCTION TRANSISTOR

**DESCRIPTION:**

The **2N2646** is a Unijunction Transistor Used in General Purpose Pulse, Timing, Sense and Trigger Applications.

**MAXIMUM RATINGS**

<b>I<sub>C</sub></b>	2.0 A (PULSED)
<b>V<sub>CE</sub></b>	30 V
<b>P<sub>DISS</sub></b>	300 mW @ T <sub>C</sub> = 25 °C
<b>T<sub>J</sub></b>	-65 °C to +125 °C
<b>T<sub>STG</sub></b>	-65 °C to +150 °C
<b>θ<sub>JC</sub></b>	33 °C/W



**CHARACTERISTICS** T<sub>C</sub> = 25 °C

SYMBOL	TEST CONDITIONS	MINIMUM	TYPICAL	MAXIMUM	UNITS
η	V <sub>B2B1</sub> = 10 V	0.56		0.75	--
r <sub>BB</sub>	V <sub>B2B1</sub> = 3.0 V	4.7		9.1	KΩ
α <sub>rBB</sub>	V <sub>B2B1</sub> = 3.0 V T <sub>A</sub> = -55 to 125 °C	0.1		0.9	%/°C
V <sub>EB1(SAT)</sub>	V <sub>B2B1</sub> = 10 V I <sub>E</sub> = 50 mA		3.0		V
I <sub>B2(MOD)</sub>	V <sub>B2B1</sub> = 10 V I <sub>E</sub> = 50 mA		20		mA
I <sub>B2EO</sub>	V <sub>B2E</sub> = 30 V I <sub>B1</sub> = 0			12	μA
I <sub>P</sub>	V <sub>B2B1</sub> = 25 V			5.0	μA
I <sub>V</sub>	V <sub>B2B1</sub> = 20 V R <sub>B2</sub> = 100 Ω	4.0			mA
V <sub>OB1</sub>	V <sub>B2B1</sub> = 20 V R <sub>B1</sub> = 20 Ω	3.0	5.0		V

ADVANCED SEMICONDUCTOR, INC.

REV. A

7525 ETHEL AVENUE • NORTH HOLLYWOOD, CA 91605 • (818) 982-1202 • TELEX: 18-2651 • FAX (818) 765-3004

1/1

2N2646




# MAC218A6FP, MAC218A10FP

Preferred Device


## Triacs

### Silicon Bidirectional Thyristors

Designed primarily for full-wave ac control applications, such as light dimmers, motor controls, heating controls and power supplies.

- Blocking Voltage to 800 Volts
- Glass Passivated Junctions for Greater Parameter Uniformity and Stability
- Isolated TO-220 Type Package for Ease of Mounting
- Gate Triggering Guaranteed in Four Modes
-  Indicates UL Registered — File #E69369
- Device Marking: Logo, Device Type, e.g., MAC218A6FP, Date Code

#### MAXIMUM RATINGS (T<sub>J</sub> = 25°C unless otherwise noted)

Rating	Symbol	Value	Unit
Peak Repetitive Off-State Voltage <sup>(1)</sup> (T <sub>J</sub> = -40 to +125°C, Sine Wave 50 to 60 Hz, Gate Open)	V <sub>DRM</sub> , V <sub>RRM</sub>	400 800	Volts
On-State RMS Current (T <sub>C</sub> = +80°C) <sup>(2)</sup> Full Cycle Sine Wave 50 to 60 Hz	I <sub>T(RMS)</sub>	8.0	Amps
Peak Non-Repetitive Surge Current (One Full Cycle, 60 Hz, T <sub>C</sub> = +80°C) Preceded and followed by rated current	I <sub>TSM</sub>	100	Amps
Circuit Fusing Considerations (t = 8.3 ms)	I <sup>2</sup> t	40	A <sup>2</sup> s
Peak Gate Power (T <sub>C</sub> = +80°C, Pulse Width = 10 μs)	P <sub>GM</sub>	16	Watts
Average Gate Power (T <sub>C</sub> = +80°C, t = 8.3 ms)	P <sub>G(AV)</sub>	0.35	Watt
Peak Gate Current (T <sub>C</sub> = +80°C, Pulse Width = 10 μs)	I <sub>GM</sub>	4.0	Amps
RMS Isolation Voltage (T <sub>A</sub> = 25°C, Relative Humidity ≤ 20%) 	V <sub>(ISO)</sub>	1500	Volts
Operating Junction Temperature	T <sub>J</sub>	-40 to +125	°C
Storage Temperature Range	T <sub>stg</sub>	-40 to +150	°C


(1) V<sub>DRM</sub> and V<sub>RRM</sub> for all types can be applied on a continuous basis. Blocking voltages shall not be tested with a constant current source such that the voltage ratings of the devices are exceeded.

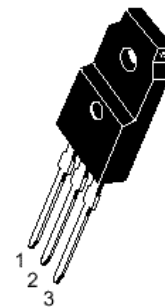
(2) The case temperature reference point for all T<sub>C</sub> measurements is a point on the center lead of the package as close as possible to the plastic body.



**ON Semiconductor**

<http://onsemi.com>

**ISOLATED TRIAC **  
**8 AMPERES RMS**  
**400 thru 800 VOLTS**



**ISOLATED TO-220 Full Pack**  
**CASE 221C**  
**STYLE 3**

PIN ASSIGNMENT	
1	Main Terminal 1
2	Main Terminal 2
3	Gate

#### ORDERING INFORMATION

Device	Package	Shipping
MAC218A6FP	ISOLATED TO220FP	500/Box
MAC218A10FP	ISOLATED TO220FP	500/Box

Preferred devices are recommended choices for future use and best overall value.

**MAC218A**

**THERMAL CHARACTERISTICS**

Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Case	$R_{\theta JC}$	2.2	°C/W
Thermal Resistance, Case to Sink	$R_{\theta CS}$	2.2 (typ)	°C/W
Thermal Resistance, Junction to Ambient	$R_{\theta JA}$	60	°C/W
Maximum Lead Temperature for Soldering Purposes 1/8" from Case for 10 Seconds	$T_L$	260	°C

**ELECTRICAL CHARACTERISTICS** ( $T_C = 25^\circ\text{C}$  unless otherwise noted; Electricals apply in both directions)

Characteristic	Symbol	Min	Typ	Max	Unit
----------------	--------	-----	-----	-----	------

**OFF CHARACTERISTICS**

Peak Repetitive Blocking Current ( $V_D = \text{Rated } V_{DRM}, V_{RRM}; \text{ Gate Open}$ )	$I_{DRM}, I_{RRM}$	—	—	10	$\mu\text{A}$
		—	—	2.0	mA

**ON CHARACTERISTICS**

Peak On-State Voltage <sup>(1)</sup> ( $I_{TM} = \pm 11.3 \text{ A Peak}$ )	$V_{TM}$	—	1.7	2.0	Volts
Gate Trigger Current (Continuous dc) ( $V_D = 12 \text{ Vdc}, R_L = 100 \Omega$ ) MT2(+), G(+) MT2(+), G(-) MT2(-), G(-) MT2(-), G(+)	$I_{GT}$	—	—	50 50 50 75	mA
Gate Trigger Voltage (Continuous dc) (Main Terminal Voltage = 12 Vdc, $R_L = 100 \text{ Ohms}$ ) MT2(+), G(+) MT2(+), G(-) MT2(-), G(-) MT2(-), G(+)	$V_{GT}$	—	0.9 0.9 1.1 1.4	2.0 2.0 2.0 2.5	Volts
Gate Non-Trigger Voltage (Continuous dc) (Main Terminal Voltage = 12 V, $R_L = 100 \Omega, T_J = +125^\circ\text{C}$ ) All Four Quadrants	$V_{GD}$	0.2	—	—	Volts
Holding Current ( $V_D = 12 \text{ Vdc}, \text{ Gate Open}, \text{ Initiating Current} = \pm 200 \text{ mA}$ )	$I_H$	—	—	50	mA

**DYNAMIC CHARACTERISTICS**

Critical Rate of Rise of Commutating Off-State Voltage ( $V_D = \text{Rated } V_{DRM}, I_{TM} = 11.3 \text{ A}, \text{ Commutating } di/dt = 4.1 \text{ A/ms}, \text{ Gate Unenergized}, T_C = 80^\circ\text{C}$ )	$dv/dt_{(c)}$	—	5.0	—	V/ $\mu\text{s}$
Critical Rate of Rise of Off-State Voltage ( $V_D = \text{Rated } V_{DRM}, \text{ Exponential Voltage Rise}, \text{ Gate Open}, T_J = 125^\circ\text{C}$ )	$dv/dt$	—	100	—	V/ $\mu\text{s}$

(1) Pulse Test: Pulse Width  $\leq 2.0 \text{ ms}$ , Duty Cycle  $\leq 2\%$ .**MAC218A**